

Elimination of interharmonic currents by a FAP using the technique of compensation global

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ABSTRACT

This document is a contribution to improve the quality of electrical energy in the distribution network. In this article we will present a method that allows us to clean up in a very considerable way the electrical network of both harmonics and inter harmonics provoked, basically by the ultrafast switches used when controlling a PWM inverter supplying a three-phase asynchronous motor. We notice that this method can be generalized for other loads creating inter-harmonics. This proposed method was simulated using the MATLAB/SIMULINK software and had given remarkable results (there is a considerable reduction in total harmonic distortion (THD) of source current from 29.52% to 0.82%)

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1. INTRODUCTION

The increasing use of the control systems based on electronic power causes more and more disturbance problems in the power grids. In fact, in addition to the harmonic disturbances, others called "inter-harmonic" perturbations (harmonics whose frequencies are non-integer multiples of the fundamental frequency) are mainly created by the non-linear loads connected to the network, especially by the converters of power (variators of speed, etc.) [1, 2].

It should be noted that interharmonics are considered to be more harmful than harmonic components because they overlap to the fundamental wave and modify not only the shape of the current, but also the voltage, which results in the following adverse effects [3-5]. Interference with telecommunication networks: The electromagnetic coupling between the electrical and telecommunication networks can induce in these last important noises. The malfunctions of some electrical equipment: In the presence of interharmonics, the voltage (or the current) can change several times of sign in half a period. The risk of resonance excitation: The resonance frequencies of the circuits formed by transformer inductances and cables are normally high; The low frequency oscillations of mechanical systems; Disturbance of fluorescent lighting and electronic equipment; Interference with control and protection signals on power supply lines; Acoustic disturbances; The saturation of the intensity transformers.

In the literature, we find that many researchers have treated the problem of harmonics generated by loads such as rectifiers, alternators, transformers, fluorescent lamps, etc. But other loads such as

asynchronous motors controlled by HF techniques can generate not only harmonics but also interharmonics [6, 7]. Several techniques have been mentioned by researchers to combat the effects of interharmonics [8]. The most used is to install passive filters that attenuate the parasitic currents before they propagate in the network. But the main disadvantage of these devices, in addition to resonance they are not adapted to the variations of interharmonic spectra [8]. The new structures of network's depollution are emerged under name active filter [3, 9]. The objective is then to adapt the filter in real time compared to the loads variations. In our work, we mainly focus on the reduction, and even the cancellation, of interharmonics resulting from the MLI control of an inverter which supplies a three-phase asynchronous motor by a parallel active filter (FAP) by using multi-filter variables (FMV) [10-12].

2. RESOLUTION METHODOLOGY

2.1. Position of the problematic

A three-phase asynchronous motor fed through a PWM inverter injects into the power grid, in addition to the harmonic currents, a lot of inter-harmonic currents that will deform the form of the voltage wave.

2.2. Proposed solution

We propose a solution that goes through two main stages.

2.2.1. Identification of interharmonic currents

To depollute the electrical network of these inter-harmonics, they must be identified with the best possible precision because the quality of the depollution resides in the efficiency of the method used for the identification of the inter-harmonic currents, several methods have been used by other researchers such as, the instantaneous power theory [6, 13, 14] the synchronous reference theory [15, 16], the theory based on the bandpass filter [17-19]. In our work, we chose the method called "global FMV" (multi-variable filter) as shown in Figure 1.

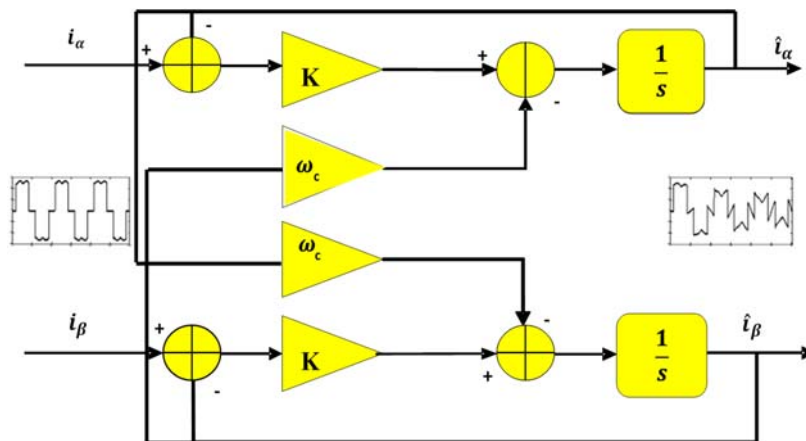


Figure 1. FMV circuit diagram

This filter essentially serves to separate the fundamental component of the load current ($f = 50 \text{ Hz}$) of harmonic and inter-harmonic components ($f \neq 50 \text{ Hz}$), noted here i_{ah} et i_{bh} , exploiting the inverse transformation of Concordia [1, 20]. Indeed, the mathematical model of the FMV is developed in the first time by Song Hong-Scok, [9] the relation between the quantities of inputs and outputs having an integral effect as the expression shows it (formula (1)) [21, 22].

$$i_{\alpha\beta}(s) = e^{j\omega_c t} \int e^{-j\omega_c t} \cdot i_{\alpha\beta}(t) dt \quad (1)$$

After the transformation of Laplace, the (1) will have the form

$$H(s) = \frac{i_{\alpha\beta}(s)}{i_{\alpha\beta}(s)} = \frac{s+jw_c}{s^2+w_c^2} \tag{2}$$

By adding two new constants k_1 and k_2 in expression (2) it will become as follows

$$H(s) = \frac{i_{\alpha\beta}(s)}{i_{\alpha\beta}(s)} = K_2 \frac{(s+K_1)+jw_c}{(s+K_1)^2+w_c^2} \tag{3}$$

With:

w_c : The cutoff pulse of the filter;

K_i : positive constants;

$i_{\alpha\beta}$: The input currents of the FMV;

$\hat{i}_{\alpha\beta}$: The output currents of the FMV;

For $H(s) = 0$ dB, it is necessary that:

$K_1 = K_2 = K$

Then

$$\hat{i}_\alpha = \frac{k}{s} [i_\alpha(s) - \hat{i}_\alpha(s)] - \frac{\omega_c}{s} \cdot \hat{i}_\beta(s) \tag{4}$$

$$\hat{i}_\beta = \frac{k}{s} [i_\beta(s) - \hat{i}_\beta(s)] - \frac{\omega_c}{s} \cdot \hat{i}_\alpha(s) \tag{5}$$

2.2.2. Compensation of interharmonic currents

Once the interharmonic currents are identified, by subtracting on each axis (α - β) the output of the FMV at its input (6) [4, 6].

$$\begin{cases} i_\alpha = \hat{i}_\alpha + i_{h\alpha} \\ i_\beta = \hat{i}_\beta + i_{h\beta} \end{cases} \tag{6}$$

The reference interharmonic currents ($i_{c_ref1,2,3}$) are obtained from the inverse transformation of Concordia

$$\begin{bmatrix} i_{c_ref1} \\ i_{c_ref2} \\ i_{c_ref3} \end{bmatrix} = \sqrt{\frac{2}{3}} \begin{bmatrix} 1 & 0 \\ -1/2 & \sqrt{3}/2 \\ -1/2 & -\sqrt{3}/2 \end{bmatrix} \begin{bmatrix} i_{ah} \\ i_{bh} \end{bmatrix} \tag{7}$$

These reference currents are used to control the voltage inverter as shown in Figure 2. The operating principle of the parallel active filter is explained in the following section.

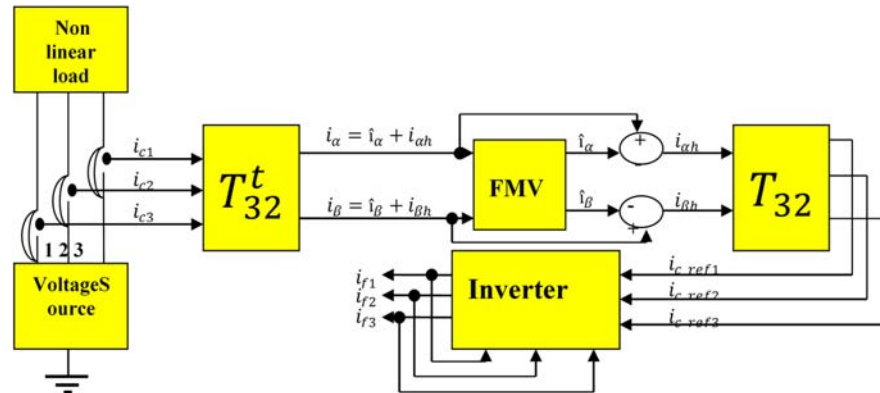


Figure 2. Algorithm for identifying inter-harmonic currents

3. PRINCIPLE OF FUNCTIONING OF A PARALLEL ACTIVE FILTER (FAP)

Our parallel active filter (FAP) is a voltage inverter with MLI; it is connected in parallel with the polluted load as shown in Figure 3. The control-command part is mainly based on the algorithm for identifying the inter-harmonic currents associated with the PWM algorithm [7, 18, 19]. Both of the

algorithms contribute to the generation of filter control signals. The regulation of the DC bus voltage is guaranteed by the classical Proportional-Integral (PI) controller. The active filter produces inter-harmonic currents which are injected into the electrical network of the same amplitude but in opposition phase with the inter-harmonic currents caused by the load; this allows to eliminate the polluting inter-harmonics and to obtain the return for the form of linear current. Reacting quickly, the active filter improves the quality of the network power supply in real time for limiting the damages [10, 23-25].

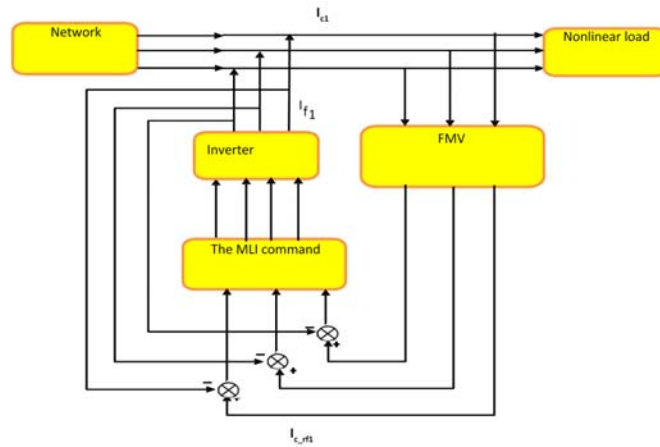


Figure 3. Principle function of the parallel active filter

4. THE ANALYSIS OF RESULTS

4.1. Source current before filtering

We note that our charge causes a significant degradation on the waveform of the current by generating harmonics and inter-harmonics of different frequencies; this degradation is translated by a high THD is worth 29.52%. On the other hand, the proliferation of such a polluted current in the electrical network has a detrimental effect on the form of the voltage, hence the need to reduce the THD to acceptable values. Figure 4 to Figure 6 show the source current before filtering, source current spectrum before filtering, and the interharmonic current.

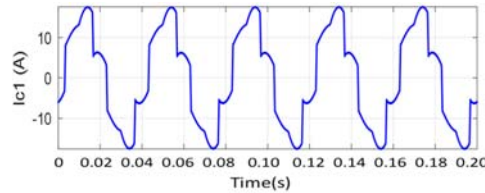


Figure 4. Source current before filtering

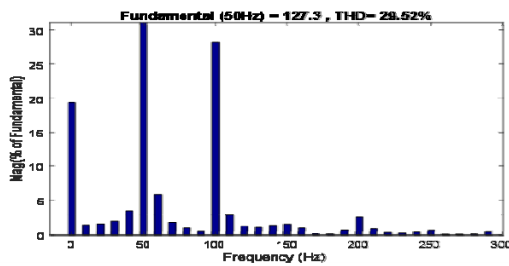


Figure 5. Source current spectrum before filtering

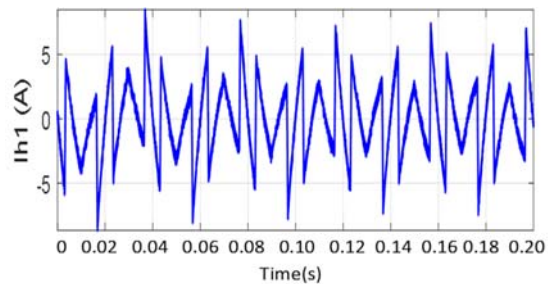


Figure 6. Interharmonic current

4.2 Source current after filtering

After the filtering, the THD becomes 0.82%, the harmonics and interharmonics took minimum values as shown in Figure 8, which confirms the efficiency of the technique used. Figure 7 shows the source current after filtering.

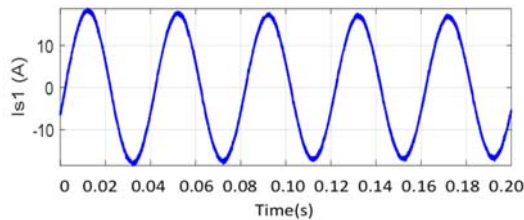


Figure 7. Source current after filtering

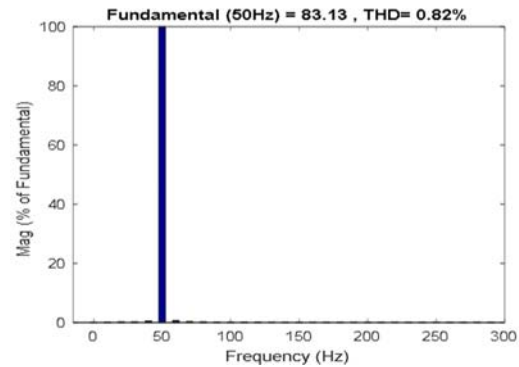


Figure 8. The source current spectrum after filtering

5. CONCLUSION

In this paper we have put in place a method that allowed us to clean up the electrical network of harmonics and interharmonics generated by a three-phase asynchronous motors controlled by HF techniques (including MLI). The results of our study can be summarized in the following points: The use of a parallel active filter employing a multi-variable filter with the technique of global extraction of harmonics and interharmonics, which allows us to depollute the almost electrical network from all kind of parasite; The FMV identification method is very effective for extracting reference interharmonics and easy to implement; The FMV technique using the compensation global allows us to completely compensate the interharmonics of the network.

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