

## H7 three phase transformerless inverter for photovoltaic grid-tied system with maximum power point operation

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### ABSTRACT

One of the most important and common parts of the modern power systems is the grid-connected photovoltaic (PV) systems. Recently, these systems have gotten a big revolution due to the introduction of the transformerless inverters. It has the merits of small size, low cost, and high efficiency. However, transformerless inverters has a general safety problem related to the earth leakage current. Various researches were directed toward evolving their performance and diminishing the leakage current to the standard limits. This article proposes an application of the H7 controller to a PV powered grid-tied three phase transformerless inverter. The transformerless inverter is linked with the grid through a boost converter. The boost converter inductance is rearranged and divided to reduce the earth leakage current of the system. simulations are carried out for the proposed H7 PV grid-tied system and for a system that uses the conventional three phase inverter. The simulation results show that the H7 inverter provides lower leakage current, higher efficiency, and lower total harmonic distortion (THD) compared to the conventional three phase inverter.

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## 1. INTRODUCTION

Solar photovoltaic (PV) energy sources have gained great importance in recent years. It is one of the most promising renewable energy sources that can meet the increasing demands of electrical energy in the future. PV sources can also overcome the problems of environmental pollution [1]. PV systems could deliver energy to electric utility grid or standalone loads. Grid tied PV systems have several advantages over standalone ones where optimum utilization of PV panels can be achieved without any storage requirement [2].

In PV grid-connected systems, a transformer is employed to achieve several targets; (1) raising the PV panels voltage to suitable levels that meet requirements of connection to the grid, (2) preventing the injection of DC currents into the grid and therefore it improves the grid power quality and (3) providing galvanic isolation between the PV panels and the grid thus achieving personal safety where it prevents the leakage current which can flow in the parasitic capacitance between the frame of PV panels and ground. On the other hand, the transformer results in increase in size, weight and cost of PV systems and also the efficiency is decreased [3]-[5]. Transformerless inverters overcome the aforementioned disadvantages of transformers. Several single phase and three phase topologies of transformerless inverters are presented in the

literature [6]-[13]. Single phase transformerless inverters are usually small-scale up to 5–6kW. Large DC link capacitors are required to overcome the pulsation in the output power. These capacitors decrease the reliability and lifetime of the system [14]. On the other hand, in transformerless inverters systems, the ac power on the output is constant, which means that large capacitors are not needed, resulting to smaller cost, and a higher reliability and longer lifetime of the whole system. Three phase transformerless inverter systems rather than single phase ones are usually employed in high power applications. Most transformerless are oriented to control the changes in common mode voltage and consequently reduce the leakage current to safe limits. The absence of maximizing the power drawn from the PV panels with reducing the ground leakage current is clearly observed [15]-[20]. The problem this paper is addressing is the leakage current minimization in three phase grid tied PV systems with achievement of maximum power operation of PV panels. H7 three phase transformerless inverter operated at SVPWM is utilized to overcome the problem of leakage current. DC-DC boost converter that is controlled by a maximum power point tracker MPPT is introduced to maximize the delivered power from PV arrays. Input voltage of the inverter is maintained constant by a voltage controller. Three phase LC filter is used to cancel high frequency harmonics. The paper is organized as follows: Section 2 introduces the techniques of ground leakage current minimization. H7 operation and its modulation techniques are presented in section 3. The overall system and controllers are proposed in section 4. Section 5 provides the simulation results and discussion of the proposed system. Section 6 provides the net conclusions of the paper.

## 2. LEAKAGE CURRENT MINIMIZATION TECHNIQUES

The ground leakage current finds its path through the parasitic capacitance ( $C_{PV}$ ) between the frame of PV panels and ground. The common mode voltage of the three-phase transformerless inverter is presented to study its influence on the leakage ground current. The German standard, VDE0126-1-1, shows the requirements for limits regarding ground leakage currents [21]. With respect to three phase transformerless inverters, the presented techniques to minimize the leakage current can be divided into two main categories. The first is based on modulation techniques and the second is based on the configurations of the topologies. The following sections introduce brief descriptions about the two categories.

### 2.1. Leakage current minimization based on modulation techniques

Reduced common mode voltage pulse width modulation (RCMV-PWM) techniques have been presented to reduce the common mode voltage and consequently reduce the ground leakage current. The most popular ones are near-state PWM (NSPWM) [22], remote-state PWM (RSPWM) [23] and active zero-state PWM (AZPWM) [24]. Unlike the conventional sinusoidal PWM (SPWM) and space vector PWM (SVPWM), the RCMV-PWM techniques do not apply the zero vectors  $V_0$  (000) or  $V_7$  (111) to the inverter operation. Instead, the RCMV-PWM techniques apply only the active vectors  $V_1$  through  $V_6$  to the inverter. The selection of active vectors over each switching period depends on the location of the voltage command in the dq plane and the type of operation (type A or type B). Figure 1 (a) and Figure 1 (b) give the active vectors and the six sectors ( $A_1$  to  $A_6$  for type A operation) and ( $B_1$  to  $B_6$  for type B operation). AZPWM and RSPWM are applied in type 'A' while NSPWM is applied in type 'B'. Table 1 summarizes the patterns of vectors during each sector when applying SVPWM and the RCMV-PWM techniques.

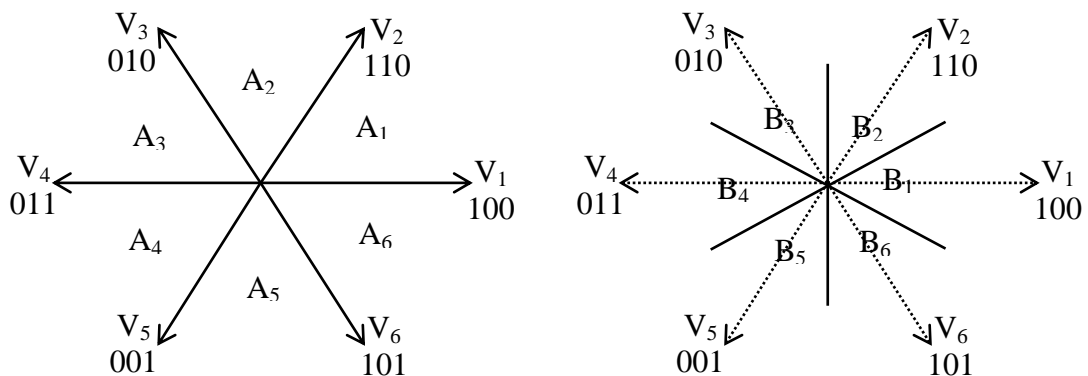


Figure 1. Voltage vectors and sectors for the two types A and B

Table 1. Patterns of vectors during each sector when applying svpwm and the RCMV-PWM

Sector	SVPWM	AZPWM	RSPWM	Sector	NSPWM
A <sub>1</sub>	7210127	6123216		B <sub>1</sub>	21612
A <sub>2</sub>	7230327	4321234		B <sub>2</sub>	32123
A <sub>3</sub>	7430347	2345432	31513 or	B <sub>3</sub>	43234
A <sub>4</sub>	7450547	6543456	42624 in	B <sub>4</sub>	54345
A <sub>5</sub>	7650567	4561654	all sectors	B <sub>5</sub>	65456
A <sub>6</sub>	7610167	2165612		B <sub>6</sub>	16561

Although the RCMV-PWM techniques reduce the ground leakage current to very small values, other system performance such as voltage linearity, output current ripples, inverter losses and total harmonic distortion are degraded. In addition, there are limitations on modulation index when applying either NSPWM or RSPWM [25]. When applying RSPWM, very high input voltage is required to inject the real power to the grid resulting in increase in losses and THD.

## 2.2. Leakage current minimization based on inverter configuration

Another approach to reduce the leakage current is based on inverter configuration. Different topologies are presented [10], [11]. Three phase full bridge VSI with split capacitor (3FB-SC) and three phase neutral point clamped VSI (3xNPC) reduce the leakage to very low values while conventional three phase full bridge VSI (3FB) has a high leakage current and then strongly requires a galvanic isolation [11]. 3FB inverter has a high leakage current when connected to grid and therefore requires a galvanic isolation. On the other hand, 3FB-SC and 3xNPC greatly reduce the leakage current below the limited values. The main disadvantage is the influence of the inductance in the neutral line which can result in leakage current higher than the allowed levels. In addition, large number of power switches and diode is required in case of 3xNPC inverter which increase the losses, system complexity and cost. Recent topologies such as three phase VSI with 4-legs, H8 converter topology and flying capacitor transformer-less inverter are presented [26]-[28] and successfully reduce the leakage current. However, the three topologies have many semiconductor elements which also increase the losses, system complexity and cost.

## 3. H7 OPERATION AND ITS MODULATION TECHNIQUE

H7 three phase transformerless inverter represents a very suitable solution to reduce the ground leakage current while adding only one switch to the conventional three phase bridge inverter as shown in Figure 2. H7 transformerless inverter is considered as extension version of well-known H5 single phase transformerless inverter. This section presents the proposed system configuration, system analysis and principle of operation of H7 three phase transformer-less inverter.

The basic idea of H7 inverter is to disconnect the PV panels from the grid during freewheeling periods and as a result there will be no path for leakage current. The seventh switch  $S_7$  conducts during active modes (vectors  $V_1$  to  $V_6$ ) and it is forced to 'off' state during freewheeling periods (vectors  $V_0$  and  $V_7$ ). A simplified equivalent circuit of H7 inverter with the grid is presented in Figure 3 including LC filter, parasitic capacitance and ground inductance  $L_g$ .

Referring to Figure 2 and during the active mode  $V_2$  (110), the switches  $S_1$ ,  $S_3$  and  $S_2$  conduct together. The inverter phase voltages and common mode voltage becomes:

$$V_{AN} = V_{BN} = \frac{V_{dc}}{3} \quad (1)$$

$$V_{CN} = \frac{-2V_{dc}}{3} \quad (2)$$

The inverter terminal voltages with respect to the negative terminal of the DC link voltage are:

$$V_{A-} = V_{B-} = V_{dc} \quad (3)$$

$$V_{C-} = \text{zero} \quad (4)$$

The common mode voltage in this case is:

$$V_{CMV} = \frac{(V_{A-} + V_{B-} + V_{C-})}{3} = \frac{2V_{dc}}{3} \quad (5)$$

The same calculations can be carried out for the other five active vectors. The results are summarized in Table 2. When applying the zero vector  $V_7$ , freewheeling operation occurs. The current paths through one of the upper switches and two antiparallel diodes of the other two switches depending on the voltage levels of the grid. During this freewheeling period:

$$V_{A-} = V_{B-} = V_{C-} = V_{dc} \quad (6)$$

and

$$V_{CMV} = V_{dc} \quad (7)$$

When applying the zero vector  $V_0$ , freewheeling operation occurs. The current passes through two switches of the lower switches and one antiparallel diode of the third switch depending on the voltage levels of the grid. During this freewheeling period:

$$V_{A-} = V_{B-} = V_{C-} = zero \quad (8)$$

and

$$V_{CMV} = zero \quad (9)$$

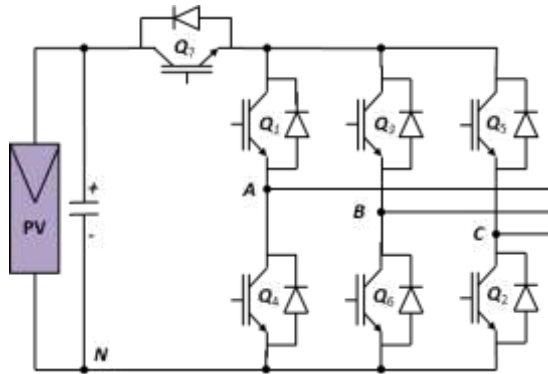


Figure 2. H7 transformerless inverter

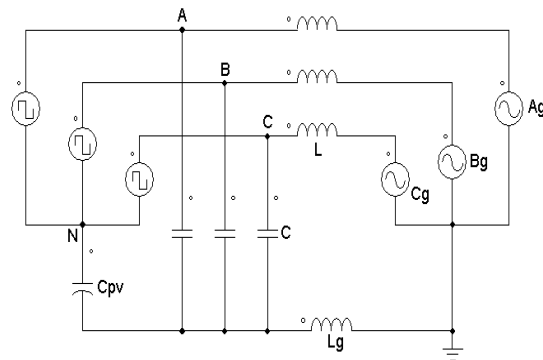


Figure 3. Simplified equivalent circuit of H7 inverter with the grid

#### 4. OVERALL SYSTEM AND CONTROLLERS

The proposed system is presented in Figure 4. The main part, H7 transformerless inverter, is the previously mentioned. The system includes DC-DC boost converter, LC filter, maximum power point tracker and controllers. The following subsections describe each of those controllers. Method of controlling the extracted power from PV panels, boost converter and H7 transformer-less inverter switches are proposed in this section.

##### 4.1. DC-DC boost converter and controller

The well-known DC-DC boost converter is utilized to raise the voltage of PV panels to meet the grid requirements. The switch of the boost converter is controlled by a maximum power point tracker to extract the maximum power from the PV panels. The inductor of the boost converter is connected in series with the PV panels the current of which is controlled through a hysteresis current controller in such a way that the PV panels' current always follow the current associated to the maximum power point at any level of sun irradiance level. A look-up table that relates the maximum power point with the corresponding current is formed at different irradiance level and is then employed to set the required reference current for the hysteresis current controller.

##### 4.2. Boost converter voltage controller

The boost converter output voltage (input voltage to the inverter) must be maintained constant (650V) in three phase system. This is achieved by a voltage controller which is presented in Figure 5 [29]. The controller combines the simplicity of classical controller and behaves like intelligent controller at the

same time. The controller compares the reference voltage (650 V) with two quantities; the actual boost voltage and its derivative multiplied by gain 1. The error is applied to gain 2 then the integrator to give the reference grid current  $I_{G-ref}$ .

**4.3. H7 inverter control signals**

Actual grid current  $I_g$ , actual grid voltage  $V_g$  and reference grid current  $I_{G-ref}$  are employed to generate the control signals of the seven switches. Phase locked loop (PLL) is applied to  $V_g$  and PI controller is used to generate the modulating voltage which is utilized to generate the high frequency control signal and its complement for each leg of the inverter. Zero voltage vectors  $V_0$  and  $V_7$  are applied during freewheeling periods as previously discussed in section 3. The phase locked loop (PLL) achieves synchronization between the PV array and the grid.

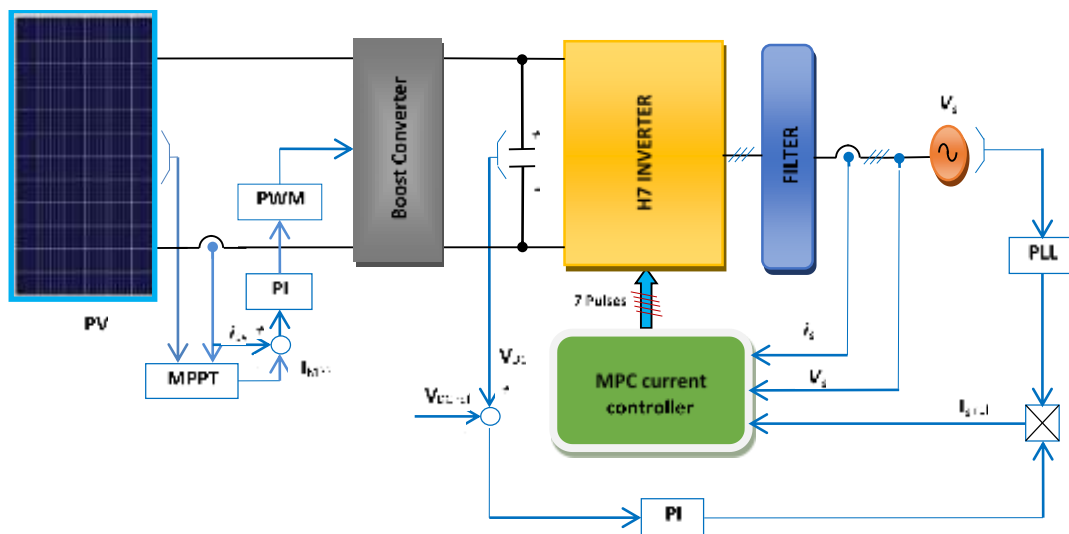


Figure 4. The proposed overall system

Table 2. Inverter phase voltages and common mode voltage when applying active vectors

	$V_1$	$V_2$	$V_3$	$V_4$	$V_5$	$V_6$
$V_{AN}$	$2V_{DC}/3$	$V_{DC}/3$	$-V_{DC}/3$	$-2V_{DC}/3$	$-V_{DC}/3$	$V_{DC}/3$
$V_{BN}$	$-V_{DC}/3$	$V_{DC}/3$	$2V_{DC}/3$	$V_{DC}/3$	$-V_{DC}/3$	$-2V_{DC}/3$
$V_{CN}$	$-V_{DC}/3$	$-2V_{DC}/3$	$-V_{DC}/3$	$V_{DC}/3$	$2V_{DC}/3$	$V_{DC}/3$
$V_{A-}$	$V_{DC}$	$V_{DC}$	0	0	0	$V_{DC}$
$V_{B-}$	0	$V_{DC}$	$V_{DC}$	$V_{DC}$	0	0
$V_{C-}$	0	0	0	$V_{DC}$	$V_{DC}$	$V_{DC}$
$V_{CMV}$	$V_{DC}/3$	$2V_{DC}/3$	$V_{DC}/3$	$2V_{DC}/3$	$V_{DC}/3$	$2V_{DC}/3$

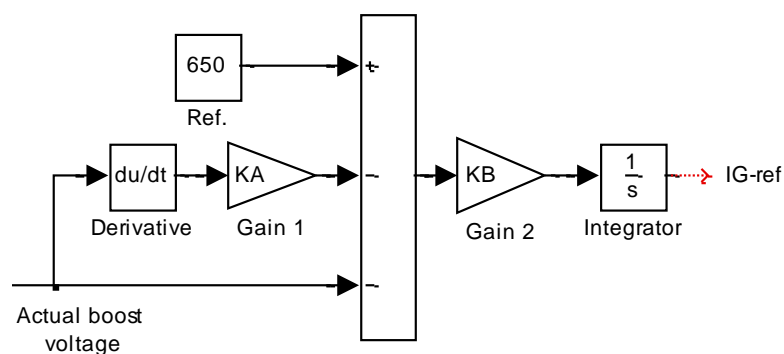


Figure 5. Boost converter voltage controller

## 5. SIMULATION RESULTS

The performance evaluation obtained through system operation and simulation results is presented in this section using MATLAB/Simulink package. The system parameters are given in Table 3. Figure 6 shows the the results of the proposed H7 transformerless inverter versus the results with the conventional 3- $\phi$  six switch inverters. Both of the topologies are controlled by PI current controller, and they operate with sinusoidal pulse width modulation SPWM. The new configuration of the boost converter is applied for each topology. The three phase grid currents are sinusoidal and in phase with the grid voltage, with the two controllers, thus (unity power factor) is achieved. The grid current THD with the H7 controller is 2.43% and that with the conventional topology is 2.45%. As the controller action in each case is different, the line voltages of inverter output have different waveforms. On the other hand, and as the same MPPT is employed for each case, the PV currents are the same for the two controllers. Based on leakage current, it can be noticed that the H7 topology case has a much smaller current than the other topology.

Figure 7 presents the variations of the RMS value of the earth leakage current as the sun irradiance level varies. In the case of the H7 topology, the leakage current value is smaller than the value of the conventional 3- $\phi$  inverter. However, at 10% insolation, vice versa occurs. The leakage current value does not exceed the standard of 300 mA for the two cases. Figure 8 gives the variations of grid current frequency spectrum of the at sun irradiance level of 100%.

The figure shows that in the H7 case, lower order harmonics are slightly higher than lower order harmonics in case of the conventional 3- $\phi$  inverter. Generally, the THD value of the two cases are very close. Figure 9 gives the response of the inverter output power and the maximum PV extracted power in the case of H7 Figure 9 (a) and the conventional 3- $\phi$  inverter Figure 9 (b). The inverter output power tracks the MPP power in the two cases with an expected steady state error which represents system losses. In the H7 controller case, the losses are smaller than that in the conventional 3- $\phi$  inverter case. This difference is due to a small THD which produces harmonic losses. With increasing power levels, losses increase. This is considered as a normal issue because current values increase with the increase in power levels. Figure 10 presents the variation of system efficiency as sun irradiance level changes for the H7 controller and conventional 3- $\phi$  inverter. Under all irradiance levels, the efficiency in case of H7 inverter is higher than that for the conventional 3- $\phi$  inverter.

Table 3. System parameters

Parameter	Value
PV short circuit current	24.53 A
PV open circuit voltage	633 V
$C_{\text{earth}}$	200 nF
DC link voltage $V_{\text{DC}}$	650 V
$L_f$	3 mH
$C_f$	2 $\mu$ F
Grid frequency	50 Hz
Grid voltage	230 V
PWM carrier frequency	9 KHz
DC link capacitor	1000 $\mu$ F

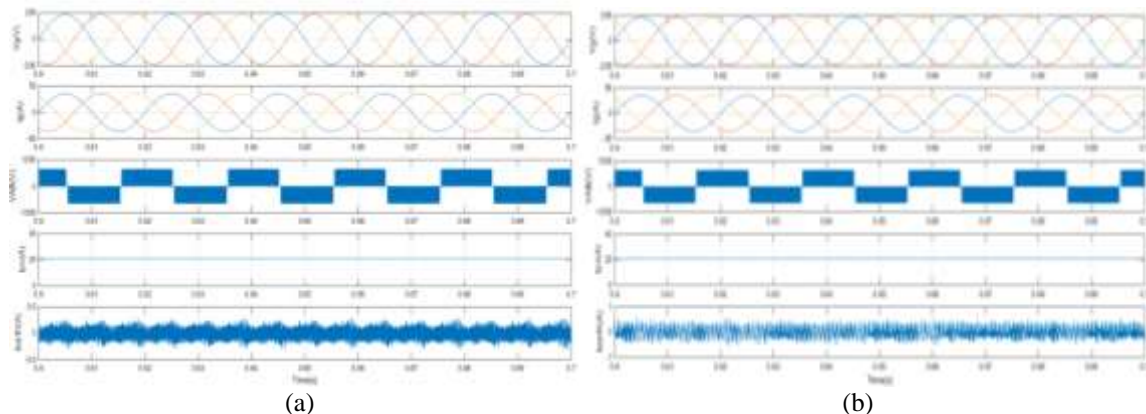


Figure 6. Simulation results of (a) the proposed H7 transformerless inverter grid voltage, grid current, PV panels' current, and ground leakage current and (b) the conventional 3- $\phi$  inverter

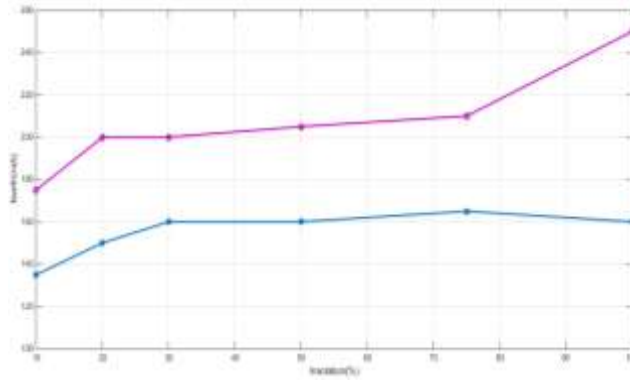


Figure 7. The variations of the ground leakage current with the sun irradiance level for the H7 inverter and the conventional 3-φ inverter

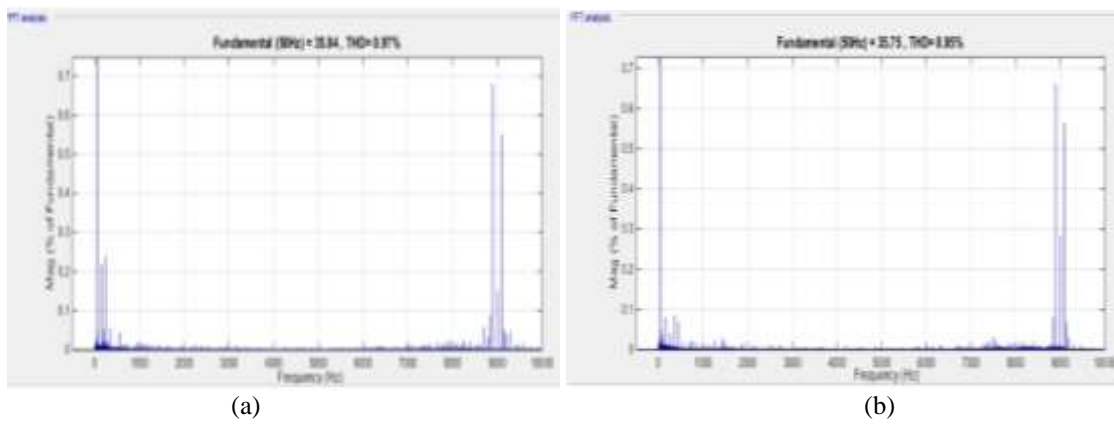


Figure 8. The grid current frequency spectrum for the (a) H7 inverter and (b) the conventional 3-φ inverter (@100% insolation)

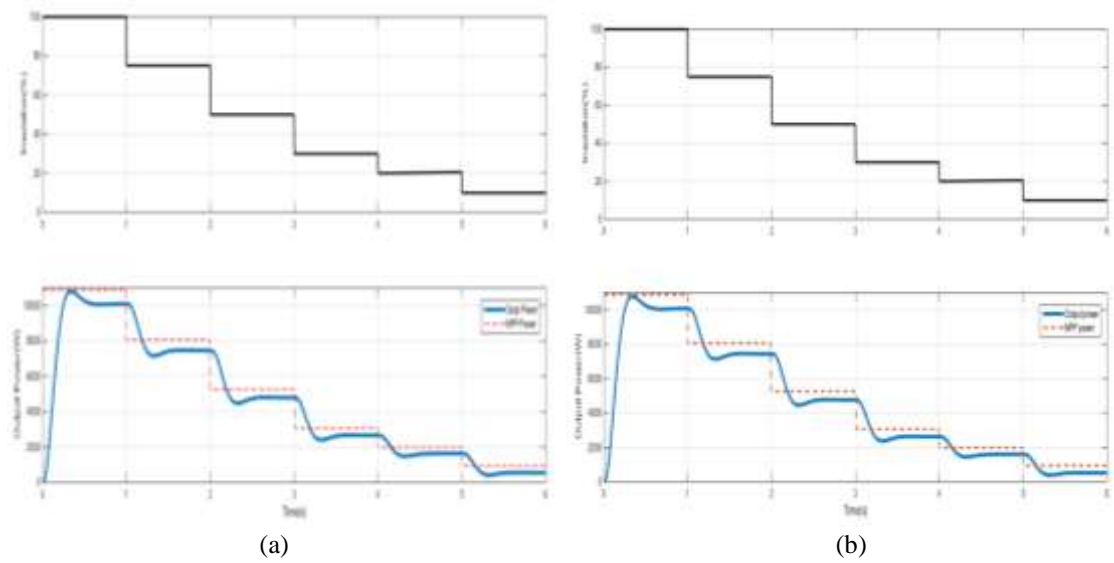


Figure 9. The MPP power, output power with variations in sun irradiance for (a) H7 topology and (b) conventional 3-φ inverter



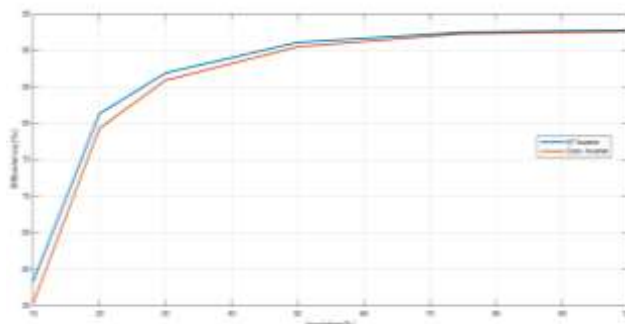


Figure 10. The system efficiency variations with variations in sun irradiance level for the H7 inverter and the conventional 3- $\phi$  inverter

## 6. CONCLUSION

This article proposed an application of the H7 controller to a PV powered grid-tied 3- $\phi$  transformerless inverter. The presented transformerless inverter is connected to the grid through a boost converter. The boost converter inductance is rearranged and divided to reduce the earth leakage current of the system. MATLAB simulations for the proposed H7 system and for a system that uses the conventional 3- $\phi$  inverter were carried out. Simulation results; i) The H7 inverter has better performance over conventional 3- $\phi$  inverter for all factors of comparison, ii) the three phase grid currents, when applying the two controllers, are sinusoidal and in phase with the utility grid voltage thus unity power factor is achieved, and iii) the THD of the grid current with the H7 inverter is 2.43% and that with conventional 3- $\phi$  inverter is 2.45%. The leakage current with the H7 inverter case is less than the value with the conventional 3- $\phi$  inverter. The efficiency of the system that uses H7 inverter is improved compared to a system that uses a conventional 3- $\phi$  inverter.

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