

# Optimal tuning fractional order PID based on marine predator algorithm for DC motor

Widi Aribowo<sup>1</sup>, Bambang Suprianto<sup>1</sup>, Reza Rahmadian<sup>1</sup>, Mahendra Widyartono<sup>1</sup>,  
Ayusta Lukita Wardani<sup>1</sup>, Aditya Prapanca<sup>2</sup>

<sup>1</sup>Department of Electrical Engineering, Faculty of Engineering, Universitas Negeri Surabaya, Surabaya, Indonesia

<sup>2</sup>Department of Computer Engineering, Faculty of Engineering, Universitas Negeri Surabaya, Surabaya, Indonesia

## Article Info

### Article history:

Received Aug 30, 2022

Revised Feb 2, 2023

Accepted Feb 17, 2023

### Keywords:

ASO

ChOA

DC motor

FOPID

Marine predator algorithm

Metaheuristic

Proportional integral derivative

STSA

## ABSTRACT

DC motors are a popular topic because they are widely applied in various electronic equipment. So, this requires a control that is fast and reliable. The development of optimized control methods is growing rapidly with the discovery of several new methods. Marine predator algorithm (MPA) is an optimization method based on marine life between living things. This article discusses the application of the MPA method for optimizing fractional order PID (FOPID) control on DC motors. The implementation of the FOPID controller is also difficult because the fractional calculus operators of the FOPID controller cannot be directly implemented in numerical calculations. The method proposed in this article is the FOPID-MPA method. To get a performance test of the proposed method, this study uses several comparison methods, namely the seagull optimization algorithm (ASO), chimp optimization algorithm (ChOA), and sine tree seed algorithm (STSA). This study also uses a variation of the reference speed to get the performance of the proposed method. From the experiment it is known that the FOPID-MPA method gives the best performance. The FOPID-MPA method has an overshoot value of 4.34% compared to the PID-MPA method in case study 1 and has an integrated of time-weighted-squared-error (ITSE) value of 7.44% better than the PID-MPA method in case study 2.

This is an open access article under the [CC BY-SA](#) license.



## Corresponding Author:

Widi Aribowo

Department of Electrical Engineering, Faculty of Engineering, Universitas Negeri Surabaya

Ketintang Campus, Surabaya 61256, East Java, Indonesia

Email: widiaribowo@unesa.ac.id

## 1. INTRODUCTION

The increase in the use of green technology for electricity generation is increasing and spreading [1]. The application of DC motors among industry or households has been so popular. DC motors are used to support electronic systems [2]–[5]. The advantages of DC motors have high torque, have no reactive power losses and do not cause harmonics in the electric power system that supplies them. PID controllers have been widely applied because of their simple structure, robustness and ease of adjustment under various operating conditions [6]–[8]. Along with the development of computing, control theory experienced a significant shift.

The fractional order controller described by a fractional order differential equation whose function is to extend the derivative and integral to a fractional order can adjust the frequency response of the control system directly and continuously. The fractional order proportional-integral-derivative controller (FOPID) has been widely applied and considerable attention in recent years from both an industrial and academic point of view. FOPID provides more flexibility in controller design compared to PID controllers [9]–[11]. FOPID has a structure of five parameters [12]. Indirectly, this implies that FOPID has a much more complex

controller setup. However, the performance of the FOPID controller also depends on the appropriate design. This is the same as the PID controller design. A well-designed FOPID controller can respond to system uncertainty and disturbance effects by providing a good dynamic response.

Methods with highly effective tuning need to be applied to replace the classic, low-performance tuning methods which have the characteristic of being time-consuming. Metaheuristic algorithm is a very popular adaptive method because it has good performance. This method has the ability to perform a stochastic search in the solution space. Several variants of the metaheuristic algorithm applied to FOPID have been presented by several researchers such as: Cuckoo search [13], [14], Harris hawks optimization [15], grasshopper optimization algorithm [16], [17], Jaya optimization algorithm [18], Henry gas solubility [19], grey wolf optimization [20]–[22], and gradient-based optimization algorithm [23]. Improved control optimization by several methods provides better performance than classic controls. From the literature reviewed, there is still a lot of room that can be exported for DC motor control in the focus of transient response.

This article presents the application of the marine predator algorithm (MPA) as an optimization of the FOPID parameter in DC motors. the contribution of this research is:

- Application of the MPA method used to tune FOPID parameters that are difficult to approach with numerical calculations.
- Comparison of MPA performance with seagull optimization algorithm (ASO), chimp optimization algorithm (ChOA), and sine tree seed algorithm (STSA).

The article is organized into the following sections: section 2 describes the MPA, DC motor and FOPID methods. In section 3, the results and analysis are reviewed. Meanwhile, section 4 contains a summary of the articles.

## 2. METHOD

### 2.1. Marine predator algorithm (MPA)

The MPA is inspired by the life of the marine ecosystem. MPA has the character that it can store the optimization results [24]. This advantage is not owned by other methods. MPA is a metaheuristic method that has characters starting from random values.

$$Y_0 = L_b + rand(U_b - L_b) \quad (1)$$

Where  $L_b$  is the lower limit and  $U_b$  is the upper limit. *rand* is a random number. Search agents in MPA are assumed to be prey and predators. The top predator called elite is found at the end of each iteration. Elite and Prey can be formulated in the following matrix:

$$Elite = \begin{bmatrix} A_{1,1}^I & A_{1,2}^I & \cdots & A_{1,d}^I \\ A_{2,1}^I & A_{2,2}^I & \cdots & A_{2,d}^I \\ \vdots & \vdots & \vdots & \vdots \\ A_{n,1}^I & A_{n,2}^I & \cdots & A_{n,d}^I \end{bmatrix} \quad (2)$$

$$Prey = \begin{bmatrix} X_{1,1}^I & X_{1,2}^I & \cdots & X_{1,j}^I \\ X_{2,1}^I & X_{2,2}^I & \cdots & X_{2,j}^I \\ \vdots & \vdots & \vdots & \vdots \\ X_{i,1}^I & X_{i,2}^I & \cdots & X_{i,j}^I \end{bmatrix} \quad (3)$$

The position of the prey is updated using the MPA's three stages. These three stages are related to the ratio of the speed of the prey to the predator.

#### i) Step 1: High speed

At this stage, prey and predators move in the same area, and this movement simulates the process of searching for prey. In this phase, the prey is looking for food and the predator is watching the movement of the prey. When  $< \frac{1}{3} \times \max\_iter$ , this movement can be formulated in (4) and (5).

$$\vec{Sh}_i = \vec{R}_b \otimes (\vec{Elite}_i - \vec{R}_b \otimes \vec{Prey}_i) \quad i = 1, 2, \dots, n \quad (4)$$

$$\vec{Prey}_i = \vec{Prey}_i + P \times \vec{R} \otimes \vec{Sh}_i \quad (5)$$

The  $\otimes$  is operation of element-wise multiplication.  $\vec{R}_b$  is a random value. It is based on Brownian motion with normal distribution.  $\vec{R} \in [0.1]$ .  $P$  is uniform random value equal to 0.5.

ii) Stage 2: Equal speed

In this phase, exploration is turned into exploitation. predators and prey have the same speed. When  $\frac{1}{3} \times \max\_iter < iter < \frac{2}{3} \times \max\_iter$ .

$$\vec{Sh}_i = \vec{R}_L \otimes (\vec{Elite}_i - \vec{R}_L \otimes \vec{Prey}_i) \quad i = 1, 2, \dots, n/2 \quad (6)$$

$$\vec{Prey}_i = \vec{Prey}_i + P \times \vec{R} \otimes \vec{Sh}_i \quad (7)$$

In the first population,  $\vec{R}_L$  denotes random numbers based on the distribution. Prey movement is simulated by  $\vec{R}_L$  multiplication. While the second half of the population, the mathematical equation is as follows:

$$\vec{Sh}_i = \vec{R}_b \otimes (\vec{R}_b \otimes \vec{Elite}_i - \vec{Prey}_i) \quad i = n/2, \dots, n \quad (8)$$

$$\vec{Prey}_i = \vec{Prey}_i + P \times CF \otimes \vec{Sh}_i \quad (9)$$

$$CF = (1 - \frac{iter}{\max\_iter})^{(2 \frac{iter}{\max\_iter})} \quad (10)$$

Predatory movements are controlled by adaptive parameters, namely  $CF$ .

iii) Stage 3: Low-speed

In this last phase, the prey has a speed below the predator. When  $iter > \frac{2}{3} \times \max\_iter$ , the mathematical equation is as follows:

$$\vec{Sh}_i = \vec{R}_L \otimes (\vec{R}_L \otimes \vec{Elite}_i - \vec{Prey}_i) \quad i = 1, \dots, n \quad (11)$$

$$\vec{Prey}_i = \vec{Prey}_i + P \times CF \otimes \vec{Sh}_i \quad (12)$$

One of the environmental problems that affect the behavior of marine ecosystems is fish aggregating devices (FADs). This is one of the equations calculated in the MPA. The FADs equation modeling is as follows:

$$\vec{Prey}_i = \begin{cases} \vec{Prey}_i + CF \times [Z_0 = Z_{min} + \vec{R} \otimes (Z_{max} - Z_{min})] \otimes A & \text{if } r \leq FADs \\ \vec{Prey}_i + [FADs(1-r) + r](\vec{Prey}_{r1} - \vec{Prey}_{r2}) & \text{if } r > FADs \end{cases} \quad (13)$$

where  $r$  is a uniform random variable.  $x_{max}$  is the upper limit and  $x_{min}$  is the lower limit. the optimization process is affected when the  $FADs$  is 0.2.  $A$  is a binary vector.

## 2.2. DC motor schematic

A DC motor is an electrical device that uses a DC input voltage as a managing parameter. Field control and armature control are controllers in DC motor. Schematically, the relationships between DC motors and armature control are shown in Figure 1.

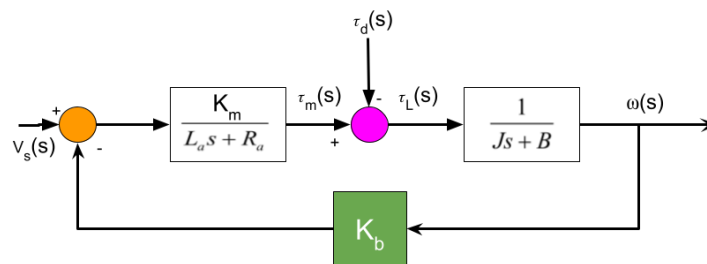


Figure 1. The DC motor schematic [25]

### 2.3. Fractional order PID

Fractional order is described by a fractional order differential which has derivatives and integrals of fractional order. The FOPID controller is a development of the PID controller model that has been around for a long time and is widely applied in industrial control systems. FOPID controls drive correction when errors between reference points and process variables occur [26]–[29]. This is done by calculating and responding with corrections that can adjust the process accordingly. The FOPID control model is as follows:

$$C_c(s) = k_p + k_i s^{-\lambda} + s^{\mu} k_d \quad (14)$$

The conventional frequency-domain method consists of three specifications, namely the crossover frequency gain ( $\omega_c$ ), phase boundary ( $\varphi_m$ ), and phase slope at  $\omega_c$ . Where  $G(s)$  is the plant and  $C(s)$  is the controller.

$$|C_c(j\omega_c)G_c(j\omega_c)| = 1 \quad (15)$$

$$\text{Arg}|C_c(j\omega_c)| + \text{Arg}|G_c(j\omega_c)| = -\pi + \varphi_m \quad (16)$$

$$\left. \frac{d|\text{Arg}|C_c(j\omega_c)G_c(j\omega_c)||}{d\omega} \right|_{\omega=\omega_c} = 0 \quad (17)$$

FOPID has five parameters to find. Where  $k_p$  is the proportional gain,  $k_i$  is the integral gain,  $k_d$  is the derivative gain,  $\lambda$  is the fractional order integral and  $\mu$  is the fractional-order derivative.  $\lambda$  and  $\mu$  are real numbers that have a range of  $0 < \lambda < 2$  and  $0 < \mu < 2$ .

### 2.4. The proposed MPA for FOPID in DC motor

DC motor control is set using MPA-based FOPID or PID Parameters. A random value is used as the initialization of MPA. MPA is limited to a predetermined iteration. DC motor control using FOPID can be illustrated in Figure 2.

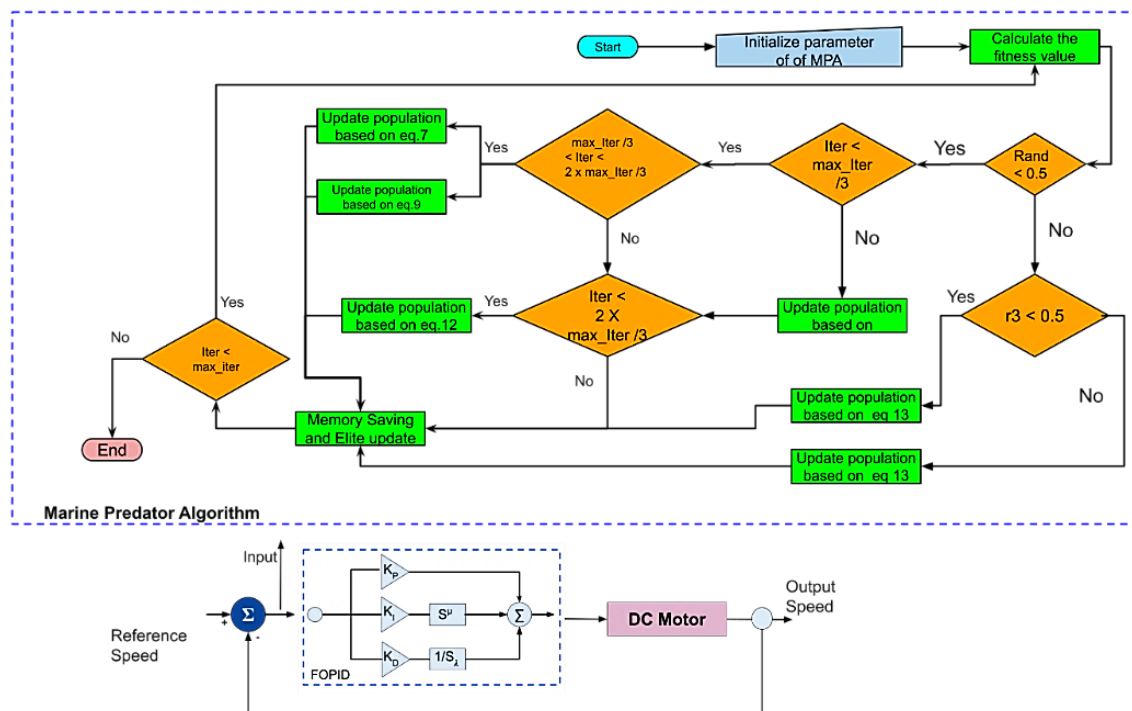


Figure 2. A FOPID based on MPA

## 3. RESULTS AND DISCUSSION

MATLAB/Simulink is used to model DC motor and implement MPA. In this paper, controller evaluation is carried out by measuring the transient response of a DC motor. FOPID parameters, namely P, I,

D,  $\lambda$  and  $\mu$  which are well regulated can produce step responses that can improve control performance. Table 1 display the MPA parameters used in this paper.

The optimal function is used to determine the performance of MPA. The convergence curve can be seen in Figure 3 (see Appendix). The unimodal function can be seen in Figures 3(a)-(g). while the multimodal function can be seen in Figures 3(h)-(m). Finally, the composite function can be seen in Figure 3(n)-(s). The optimization results of each algorithm can be seen in Table 2. Integrated of time-weighted-squared-error (ITSE) is popularly used for measuring control performance because they can perform good evaluations. The ITSE is used more as an objective function to set control gains. ITSE is most often applied to systems that require fast setup times. In this article ITSE is used as an evaluation. ITSE equation [30] is as follows:

$$ITSE = \int_0^{\infty} t \cdot e^2(t) \cdot dt \quad (18)$$

Tabel 1. MPA Parameters

No.	Parameter	Value
1	Prey	50
2	FADs	0.2
3	Iterasi	50
4	P	0.2
5	Dim	4

Tabel 2. Optimized parameters

Method	$K_p$	$K_I$	$K_d$	$\lambda$	$\mu$
PID – ChOA	3.2061	10	0.1699	-	-
PID – SOA	3.1251	10	0	-	-
PID – STSA	3.2522	10	0	-	-
PID - MPA	3.1837	10	2.4775	-	-
FOPID -MPA	3.7567	9.99	0.5422	1.0001	0.5

In this article, two case studies are used, namely study 1 with a reference speed of 1 and study 2 with variable reference speed. In case study 1, the FOPID-MPA method has the best ITSE value of 0.2779. FOPID-MPA method has an overshoot value of 1.0001. This value is close to the reference value, namely 1. In case study 1, the worst overshoot value is owned by the PID-ChOA method, which is 1.0022. The simulation results from study 1 can be seen in detail in Table 3 and Figure 4.

Table 3. DC motor output with reference speed of 1

Controller	Overshoot	Rise time	Settling time	ITSE
PID - SOA	1.0037	0.1759	0.2766	0.2939
PID-ChOA	1.0045	0.1826	0.2788	0.2958
PID-STSA	1.0027	0.1774	0.2841	0.2905
PID-MPA	1.0027	0.1774	0.2841	0.2905
FOPID-MPA	1.0019	0.1969	0.3163	0.2779

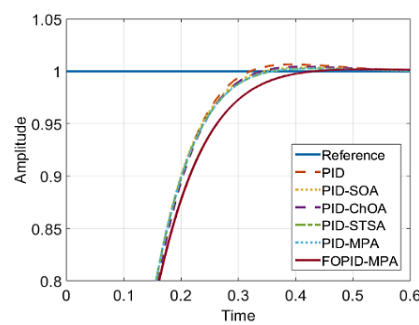


Figure 4. The result of DC motor in study 1

In case study 2, the reference speed was varied 3 times. First start, the reference speed is 0.5. Then it is increased until it reaches the reference speed of 1. Finally, the reference speed is lowered to 0.25. From the change in reference speed, the FOPID-MPA method has the best ITSE value of 0.4274. When the reference speed is 0.5, the overshoot value of the FOPID-MPA method is 0.5009. Then the reference speed is increased to 1, the FOPID-PID method has an overshoot of 1.0001. The decrease in reference speed to 0.25 was responded well by the FOPID-MPA method with an undershoot value of 0.2486. This value has an error of 0.8%. The simulation results from study 2 can be seen in detail in Table 4.

Table 4. DC motor output with variable reference speed

Controller	Overshoot step 1	Overshoot step 2	Undershoot step 3	Rise time	Settling time	ITSE
PID - SOA	0.5018	1.0018	0.2472	0.0621	1.276	0.4638
PID - ChOA	0.5022	1.0022	0.2467	0.0662	1.279	0.4565
PID - STSA	0.5014	1.0014	0.2481	0.06158	1.282	0.4592
PID - MPA	0.5013	1.0014	0.2480	0.06158	1.281	0.4592
FOPID - MPA	0.5009	1.0001	0.2486	0.0596	1.317	0.4274

#### 4. CONCLUSION

The article review usage of the MPA to FOPID in DC motors. The MPA is inspired by marine ecosystem life and has the advantage of being able to store optimization results. This article uses several comparison methods, namely ASO, ChOA, and STSA. The comparison method is used for proportional-integral-derivative (PID) control. In this article, 2 case studies are used. In case study 1, the FOPID-MPA method has the best ITSE value of 0.2779. The ITSE value of FOPID-MPA is 4.34% better than the PID-MPA method. The overshoot value of the FOPID-MPA method is 1.0019. The overshoot value of FOPID-MPA is 4.34% better than the PID-MPA method. In case study 2, the ITSE of the FOPID-MPA method is 7.44% better than the PID-MPA. In this article, it is found that the application of the FOPID-MPA method shows better performance than the PID-MPA method.

#### APPENDIX

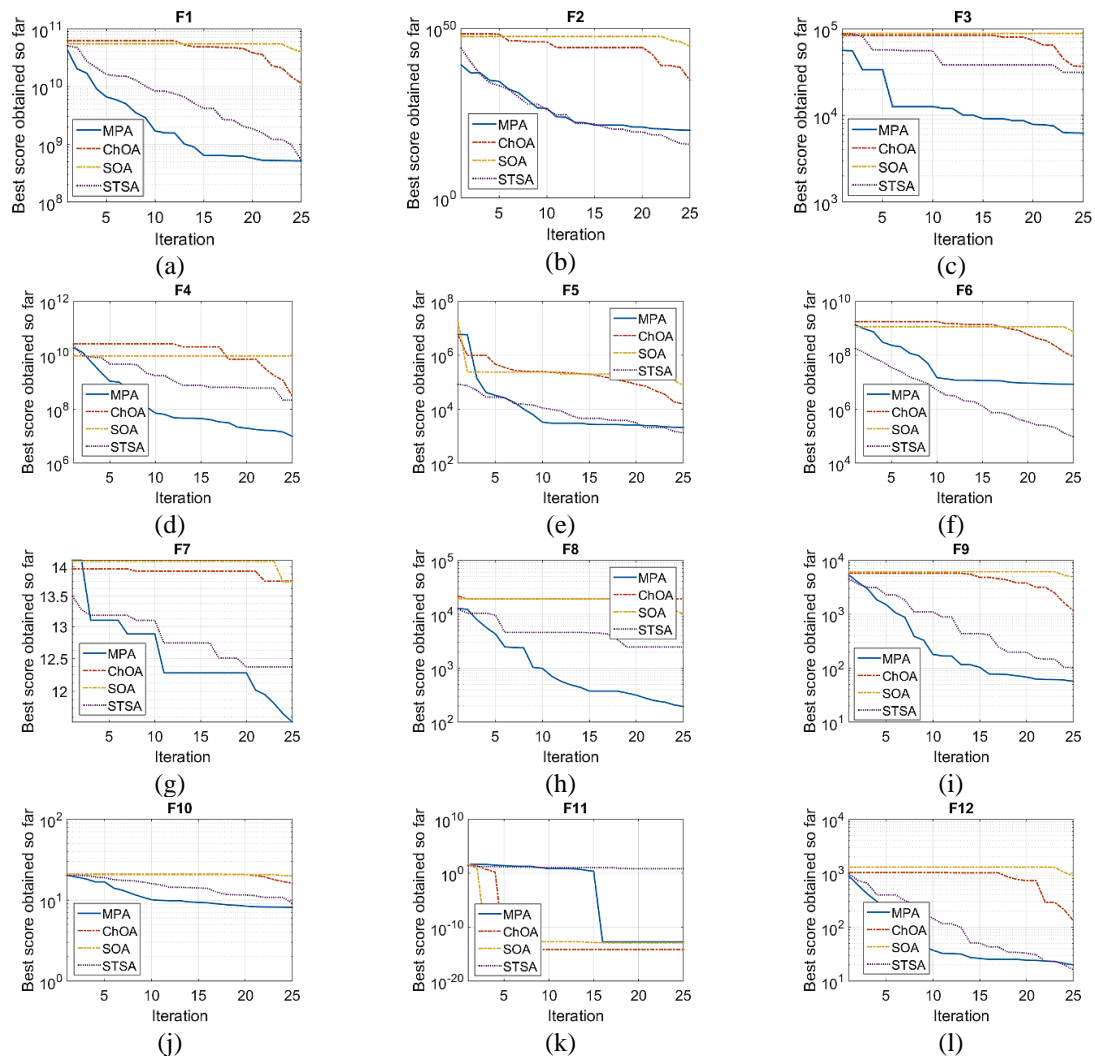


Figure 3. The convergence curve: (a) F1, (b) F2, (c) F3, (d) F4, (e) F5, (f) F6, (g) F7, (h) F8, (i) F9, (j) F10, (k) F11, and (l) F12

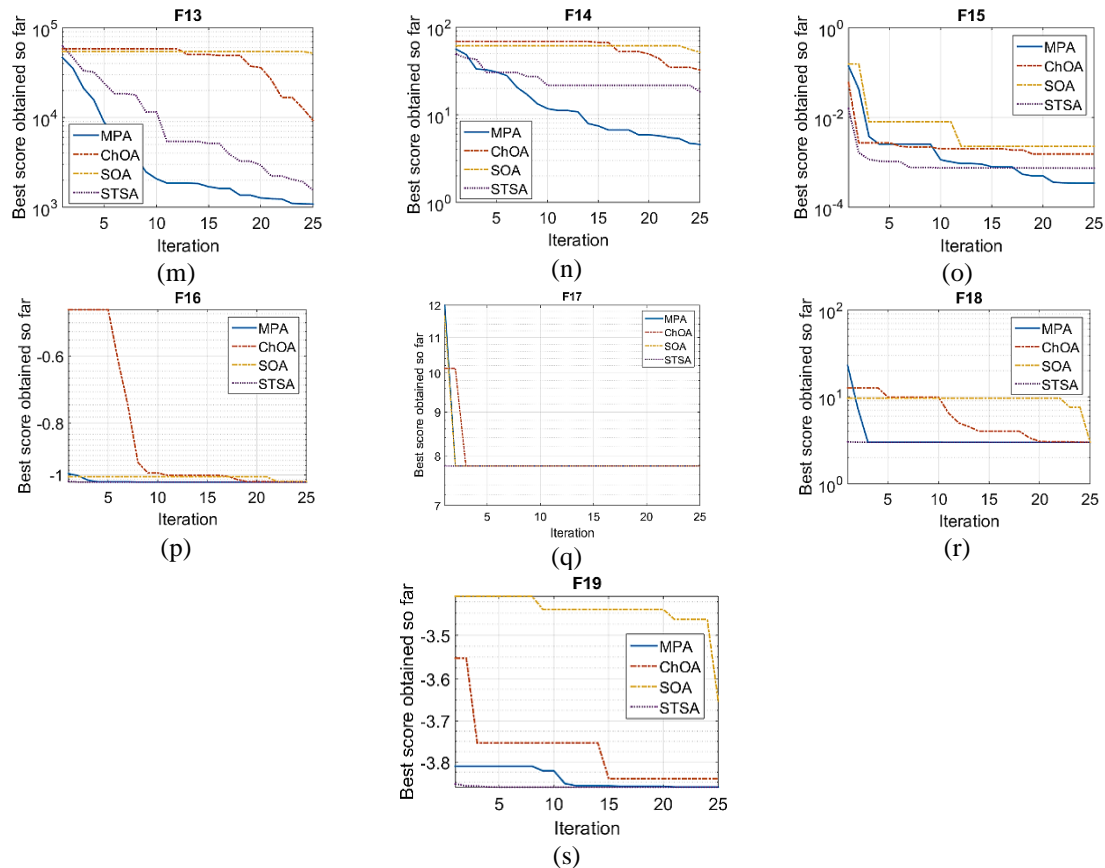


Figure 3. The convergence curve: (m) F13, (n) F14, (o) F15, (p) F16, (q) F17, (r) F18, and (s) F19 (continue)

## REFERENCES




- [1] M. Sharma, S. Dhundhara, Y. Arya, and S. Prakash, "Frequency excursion mitigation strategy using a novel COA optimised fuzzy controller in wind integrated power systems," *IET Renew. Power Gener.*, vol. 14, no. 19, pp. 4071–4085, 2020, doi: 10.1049/iet-rpg.2020.0882.
- [2] A. Ma'arif, and N. R. Setiawan, "Control of DC motor using integral state feedback and comparison with PID: simulation and arduino implementation," *J. of Robot. Control*, vol. 2, no. 5, pp. 456–461, 2021, doi: 10.18196/jrc.25122.
- [3] E. Eker, M. Kayri, S. Ekinci, and D. Izci, "A new fusion of ASO with SA algorithm and its applications to MLP training and DC motor speed control," *Arab. J. Sci. Eng.*, vol. 46, no. 4, pp. 3889–3911, 2021, doi: 10.1007/s13369-020-05228-5.
- [4] R. Silva-Ortigoza *et al.*, "Sensorless tracking control for a 'full-bridge buck inverter–DC motor' system: passivity and flatness-based design," *IEEE Access*, vol. 9, pp. 132191–132204, 2021, doi: 10.1109/ACCESS.2021.3112575.
- [5] E. Can, "A flexible closed-loop (fcl) pid and dynamic fuzzy logic+ pid controllers for optimization of dc motor," *J. Eng. Res.*, 2021, doi: 10.36909/jer.13813.
- [6] B. B. Acharya, S. Dhakal, A. Bhattacharai, and N. Bhattacharai, "PID speed control of DC motor using meta-heuristic algorithms," *Int. J. Power Electron. Drive Syst.*, vol. 12, no. 2, p. 822, Jun. 2021, doi: 10.11591/ijpeds.v12.i2.pp822-831.
- [7] W. Aribowo, Supari, and B. Suprianto, "Optimization of PID parameters for controlling DC motor based on the aquila optimizer algorithm," *Int. J. Power Electron. Drive Syst.*, vol. 13, no. 1, pp. 2808–2814, 2022, doi: ijpeds.v13.i1.pp216-222.
- [8] E. Çelik, N. Öztürk, Y. Arya, and C. Ocak, "(1 + PD)-PID cascade controller design for performance betterment of load frequency control in diverse electric power systems," *Neural Comput. Appl.*, 2021, doi: 10.1007/s00521-021-06168-3.
- [9] M. S. Ayas, and E. Sahin, "FOPID controller with fractional filter for an automatic voltage regulator," *Comput. Electr. Eng.*, vol. 90, p. 106895, 2021, doi: 10.1016/j.compeleceng.2020.106895.
- [10] M. Saadatmand, G. B. Gharehpetian, P. Siano, and H. H. Alhelou, "PMU-based FOPID controller of large-scale wind-PV farms for LFO damping in smart grid," *IEEE Access*, vol. 9, pp. 94953–94969, 2021, doi: 10.1109/ACCESS.2021.3094170.
- [11] S. V. Devaraj *et al.*, "Robust queen bee assisted genetic algorithm (QBGA) optimized fractional order PID (FOPID) controller for not necessarily minimum phase power converters," *IEEE Access*, vol. 9, pp. 93331–93337, 2021, doi: 10.1109/ACCESS.2021.3092215.
- [12] N. Khanduja, and B. Bhushan, "Optimal design of FOPID controller for the control of CSTR by using a novel hybrid metaheuristic algorithm," *Sādhanā*, vol. 46, no. 2, pp. 1–12, 2021, doi: 10.1007/s12046-021-01632-1.
- [13] M. A. Al-Gabalawy, N. S. Hosny, and S. A. Hussien, "Cuckoo search algorithm based for tuning both PI and FOPID controllers for the DFIG-Wind energy conversion system," *Int. J. Electr. Comput. Eng.*, vol. 10, no. 6, pp. 6319–6329, 2020, doi: 10.11591/ijece.v10i6.pp6319-6329.
- [14] T. Varshney, V. S. Bhadoria, P. Sonwane, and N. Singh, "Optimization of fractional-order PID controller (FOPID) using cuckoo search," in *Proceedings of 2nd International Conference on Artificial Intelligence: Advances and Applications*, 2022, pp. 649–657, doi: 10.1007/978-981-16-6332-1\_53.






- [15] V. K. Munagala, and R. K. Jatoth, "Design of fractional-order PID/PID controller for speed control of DC motor using Harris Hawks optimization," in *Intelligent algorithms for analysis and control of dynamical systems*, Springer, 2021, pp. 103–113, doi: 10.1007/978-981-15-8045-1\_11.
- [16] K. Aseem, and S. Selva Kumar, "Hybrid k-means grasshopper optimization algorithm based FOPID controller with feed forward DC–DC converter for solar-wind generating system," *J. Ambient Intell. Humaniz. Comput.*, vol. 13, no. 5, pp. 2439–2462, 2022, doi: 10.1007/s12652-021-03173-1.
- [17] B. Hekimoğlu, "Robust fractional order PID stabilizer design for multi-machine power system using grasshopper optimization algorithm," *J. Fac. Eng. Archit. Gazi Univ.*, vol. 35, no. 1, pp. 165–180, 2020, doi: 10.17341/gazimmfd.449685.
- [18] T. A. Jumani *et al.*, "Jaya optimization algorithm for transient response and stability enhancement of a fractional-order PID based automatic voltage regulator system," *Alexandria Eng. J.*, vol. 59, no. 4, pp. 2429–2440, 2020, doi: 10.1016/j.aej.2020.03.005.
- [19] S. Ekinici, D. Izci, and B. Hekimoğlu, "Henry gas solubility optimization algorithm based FOPID controller design for automatic voltage regulator," in *2020 International Conference on Electrical, Communication, and Computer Engineering (ICECCE)*, 2020, pp. 1–6, doi: 10.1109/ICECCE49384.2020.9179406.
- [20] S. K. Verma, S. Yadav, and S. K. Nagar, "Optimization of fractional order PID controller using grey wolf optimizer," *J. Control. Autom. Electr. Syst.*, vol. 28, no. 3, pp. 314–322, 2017, doi: 10.1007/s40313-017-0305-3.
- [21] J. Agarwal, G. Parmar, R. Gupta, and A. Sikander, "Analysis of grey wolf optimizer based fractional order PID controller in speed control of DC motor," *Microsyst. Technol.*, vol. 24, no. 12, pp. 4997–5006, 2018, doi: 10.1007/s00542-018-3920-4.
- [22] M. Y. Silaa, O. Barambones, M. Derbeli, C. Napole, and A. Bencherif, "Fractional order PID design for a proton exchange membrane fuel cell system using an extended grey wolf optimizer," *Processes*, vol. 10, no. 3, p. 450, 2022, doi: 10.3390/pr10030450.
- [23] S. M. A. Altbawi, A. S. Bin Mokhtar, T. A. Jumani, I. Khan, N. N. Hamadneh, and A. Khan, "Optimal design of Fractional order PID controller based automatic voltage regulator system using gradient-based optimization algorithm," *J. King Saud Univ. Sci.*, 2021, doi: 10.1016/j.jksues.2021.07.009.
- [24] A. Faramarzi, M. Heidarinejad, S. Mirjalili, and A. H. Gandomi, "Marine predators algorithm: a nature-inspired metaheuristic," *Expert Syst. Appl.*, vol. 152, p. 113377, 2020, doi: 10.1016/j.eswa.2020.113377.
- [25] W. Aribowo, S. Muslim, F. Achmad, and A. C. Hermawan, "Improving neural network based on seagull optimization algorithm for controlling DC motor," *J. Elektron. dan Telekomun.*, vol. 21, no. 1, p. 48, 2021, doi: 10.14203/jet.v21.48-54.
- [26] L. Liu, L. Zhang, G. Pan, and S. Zhang, "Robust yaw control of autonomous underwater vehicle based on fractional-order PID controller," *Ocean Eng.*, vol. 257, p. 111493, 2022, doi: 10.1016/j.oceaneng.2022.111493.
- [27] A. A. Jamil, W. F. Tu, S. W. Ali, Y. Terriche, and J. M. Guerrero, "Fractional-order PID controllers for temperature control: a review," *Energies*, vol. 15, no. 10, p. 3800, 2022, doi: 10.3390/en15103800.
- [28] L. F. da S. C. Pereira, E. Batista, M. A. G. de Brito, and R. B. Godoy, "A robustness analysis of a fuzzy fractional order PID controller based on genetic algorithm for a DC-DC boost converter," *Electronics*, vol. 11, no. 12, p. 1894, 2022, doi: 10.3390/electronics11121894.
- [29] K. Singh and Y. Arya, "Tidal turbine support in microgrid frequency regulation through novel cascade Fuzzy-FOPID droop in de-loaded region," *ISA Trans.*, 2022, doi: 10.1016/j.isatra.2022.07.010.
- [30] W. Aribowo, R. Rahmadian, M. Widyartono, A. C. Hermawan, A. L. Wardani, and B. Suprianto, "Marine predators algorithm for tuning DC motor," in *2022 Fifth International Conference on Vocational Education and Electrical Engineering (ICVEE)*, 2022, pp. 136–140, doi: 10.1109/ICVEE57061.2022.9930361.

## BIOGRAPHIES OF AUTHORS






**Widi Aribowo**    is a lecturer in the Department of Electrical Engineering, Universitas Negeri Surabaya, Indonesia. He is B.Sc in Power Engineering, Sepuluh Nopember Institute of Technology (ITS), Surabaya in 2005. He is M.Eng in Power Engineering, Sepuluh Nopember Institute of Technology (ITS), Surabaya in 2009. He is mainly research in the power system and control. He can be contacted at email: widiaribowo@unesa.ac.id.






**Bambang Suprianto**    is a lecturer in the Department of Electrical Engineering, Universitas Negeri Surabaya, Indonesia. He completed Bachelor of Electronic Engineering Education in Universitas Negeri Surabaya-Surabaya in 1986. He holds Master Engineering in Sepuluh Nopember Institute of Technology (ITS), Surabaya in 2001. He was completed Doctor of Electrical Engineering in Sepuluh Nopember Institute of Technology (ITS), Surabaya in 2012. His research interests including power system, control and electronic. He can be contacted at email: bambangsuprianto@unesa.ac.id.








**Reza Rahmadian**    received his Bachelor of Applied Science from Electronic Engineering Polytechnic Institute of Surabaya (PENS), Surabaya, Indonesia, in 2006, and his Master of Engineering Science from Curtin University, Australia, in 2013. He is currently a lecturer at the Department of Electrical Engineering, Universitas Negeri Surabaya, Indonesia. His research interests include renewable energy. He can be contacted at email: rezarahmadian@unesa.ac.id.






**Mahendra Widyartono**    received his Bachelor of Engineering from Sepuluh Nopember Institute of Technology (ITS), Surabaya, Indonesia, in 2006, and his Master of Engineering from Brawijaya University, Indonesia, in 2012. He is currently a lecturer at the Department of Electrical Engineering, Universitas Negeri Surabaya, Indonesia. His research interests include power system and renewable energy. He can be contacted at email: mahendrawidyartono@unesa.ac.id.



**Ayusta Lukita Wardani**    received her Bachelor of Applied Science from Electronic Engineering Polytechnic Institute of Surabaya (PENS), Surabaya, Indonesia, in 2011, and her Master of Engineering from Sepuluh Nopember Institute of Technology (ITS), Indonesia, in 2017. She is currently a lecturer at the Department of Electrical Engineering, Universitas Negeri Surabaya, Indonesia. Her research interests include renewable energy. She can be contacted at email: ayustawardani@unesa.ac.id.



**Aditya Prapanca**    received his Bachelor of Engineering from Sepuluh Nopember Institute of Technology (ITS), Surabaya, Indonesia, in 2000, and his Master of Computer from Sepuluh Nopember Institute of Technology (ITS), Indonesia, in 2007. He is currently a lecturer at the Department of Computer Engineering, Universitas Negeri Surabaya, Indonesia. His research interests include artificial intelligence. He can be contacted at email: adityaprapanca@unesa.ac.id.