

ANFIS based sliding mode control of a DFIG wind turbine excited by an indirect matrix converter

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ABSTRACT

Doubly fed induction generator (DFIG) is very popular in commercial turbines installed worldwide. Indirect matrix converter (IMC) is a fully silicon-based converter without capacitor that could be used in wind turbines. In this paper a robust ANFIS based second order sliding mode controller is proposed to control the DFIG wind turbine excited by IMC converter. ANFIS is a combination of fuzzy logic control and artificial neural networks. This hybrid combination could decrease the complexity of the wind turbine system. This scheme does not need the exact mathematical model of the system as needed in classical control methods. The effectiveness of the proposed controller is verified by simulation results compared to the classical PI controller.

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1. INTRODUCTION

Wind energy is growing fast throughout the world as an attractive renewable resource of electricity energy. The manufacturing technology of the wind turbines improved during the last decade, especially in power electronics section and recyclable blades technology. The rate of installed capacity of wind turbines is about 30 percent worldwide. Nowadays, the variable speed technology is dominant in wind turbine industry worldwide due to their advantages over the initial fixed speed counterpart. The doubly fed induction generator (DFIG) is the most utilized generator because of variable speed operation and low-cost power electronics drive. Schematic of a DFIG wind turbine is drawn in Figure 1. In this structure, the stator winding connection to the utility is done directly or by a transformer and the rotor three phase winding connection to the utility is accomplished by a power electronics converter. The B2B converters are used commonly to control the generator. The bulky dc-link capacitor is a challenge in this structure in case of reducing the reliability of the overall system and increasing the cost of the converter [1].

In this paper, the B2B converter structure is substituted by an indirect matrix converter (IMC) to excite the generator. The IMC's reliability and robustness are more than B2B converter due to removing the bulky electrolytic capacitor. Also, the size and weight of the IMC are less than B2B counterpart. Furthermore, the power factor in the input side is controllable in IMC and the input currents with output voltage are nearly sinusoidal with just high frequency switching harmonics. Consequently, the IMC could be utilized in wind turbines instead of traditional B2B converter due to aforementioned characteristics.

Classical control methods are investigated in some previous published papers in case of DFIG wind turbines [2]–[8]. Based on exponential reaching law a sliding mode control is proposed in [9]. The presented scheme at first control the electromagnetic torque and power factor to achieve maximum power point tracking (MPPT) in different speeds of the wind. Secondly, the chattering phenomenon and reaching speed of the states are approved in this scheme. Chattering near the sliding surface is inevitable due to high-speed switching law. Therefore, it is a challenge in case of using this controller in commercial wind turbines to avoid high frequency dynamics of the system excited by chattering of the controller. A set of adaptive and nonlinear algorithms to control variable speed wind turbines is presented in [10]. Both mechanical and electrical dynamics of the system is considered to design the controller. Validation of the presented control method is done by simulation results in capturing maximum power available from the wind and smooth speed control of the generator in different scenarios. A direct control of active-reactive powers of the generator is proposed in [11]. This method is relied on nonlinear sliding mode control to calculate the rotor voltage in order to set the active power and reactive errors of the generator to zero. Consequently, this control method does not require the internal current controller and synchronous reference frame transformation that leads to a simple control method.

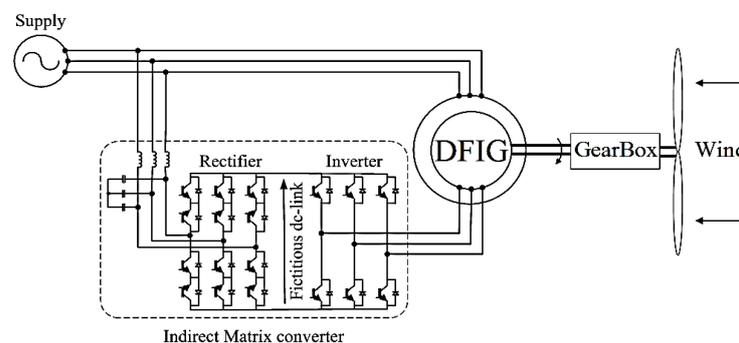


Figure 1. The IMC controlled DFIG wind turbine

The synchronizing of DFIGs without PLL to a distorted and unbalanced utility based on SMC is proposed [12]. The authors claimed that it could be avoided to mass calculations of positive and negative components by implementing in the original reference frame. The aim of this work is to synchronize the DFIG to the grid as mentioned. The behavior of the DFIG during fault based on sliding mode control is discussed in [13]. Modelling of the DFIG is done in a frame that the d-axis is aligned with the flux of the stator in this work. The stability of the system is proved by Lyapunov function. Model predictive control scheme is presented to control the DFIG wind turbine in [14]. To achieve the machine side converter control method, the model of the DFIG is linearized. So, this scheme could not be effective through wide operating area of the system and nonlinear behavior of the wind turbine for all conditions. Utilizing classical PI controller is well-known in the literature and discussed with details in [15]–[18].

Due to the merits of the matrix converters over the B2B structure, some authors investigated replacement of the B2B with the matrix converters [19]–[25]. Controlling the doubly-fed induction generators by the DTC and fuzzy logic utilizing IMC are proposed in [22], [23]. MPC Control of the DFIG using IMC is presented in [24]. The DFIG's dynamic performance utilizing IMC during grid fault has been investigated in [25]. In [26] an IMC is utilized driving a fixed speed WT. Simulation results of a DFIG wind turbine controlled by an IMC in [22] show that the harmonics content of the system is reduced.

Adaptive-neuro-fuzzy inference system based on sliding-mode-controller (ANFIS-SMC) is an interesting controller that could be used in wind turbine applications. ANFIS is a combination of fuzzy logic control and artificial neural networks. This hybrid combination could decrease the complexity of the wind turbine system. The exact model of the system is not required in the ANFIS; therefore, the realization of the control is simpler than the methods designed based on exact model. Furthermore, ANFIS based on Sugeno could bring better results compared with classical methods. Therefore, in this paper this controller is utilized for improving the dynamic response of a wind turbine system.

The remaining of the paper is categorized as follow. Control and modelling of the doubly fed induction generator is investigated in the next section. In section 3, the ANFIS-SMC control of the generator is presented in section 3. Simulation results in different wind speeds are presented in section 4. A brief conclusion of the given results is brought in the final section of the paper.

2. MODELING AND CONTROL OF THE SYSTEM

The dynamic model of a doubly fed induction generator is presented in this section. Connection of the stator winding of the DFIG to the utility is done directly or through a transformer, and the rotor winding connection to the grid is done by an IMC. The IMC rating is about 30 percent of the nominal power of the wind turbine due to this fact that the rotor winding must handle the slip power in this structure. The rotor’s shaft of the DFIG is connected to the blades’ shaft through a multi-stage gearbox that increase the low speed of the wind to the operation speed required for the DFIG.

2.1. DFIG model

The dynamic model of the DFIG is presented by a fifth-order system. The d-q components of electrical parameters of the DFIG in a reference frame that the d-axis is aligned to the flux of the stator and rotate with synchronous speed of the utility are as follows:

$$\begin{aligned}
 v_{ds} &= R_s i_{ds} + \frac{d\varphi_{ds}}{dt} - \omega_s \varphi_{qs} & \varphi_{ds} &= L_m i_{dr} + L_s i_{ds} \\
 v_{qs} &= R_s i_{qs} + \frac{d\varphi_{qs}}{dt} + \omega_s \varphi_{ds} & \varphi_{qs} &= L_m i_{qr} + L_s i_{qs} \\
 v_{dr} &= R_r i_{dr} + \frac{d\varphi_{dr}}{dt} - (\omega_s - \omega_r) \varphi_{qr} & \varphi_{dr} &= L_r i_{dr} + L_m i_{ds} \\
 v_{qr} &= R_r i_{qr} + \frac{d\varphi_{qr}}{dt} + (\omega_s - \omega_r) \varphi_{dr} & \varphi_{qr} &= L_r i_{qr} + L_m i_{qs} \\
 T_{em} &= P(\varphi_{qr} i_{dr} - \varphi_{dr} i_{qr}) & L_s &= L_m + L_{ls} \quad , \quad L_r = L_m + L_{lr}
 \end{aligned}
 \tag{1}$$

In these relations, L_s , L_r , L_{ls} , L_{lr} , R_s and R_r are the inductances and resistances of the rotor and stator windings, L_m is the magnetizing inductance, $(v_{ds}, v_{qs}, v_{dr}, v_{qr})$, $(i_{ds}, i_{qs}, i_{dr}, i_{qr})$, $(\varphi_{ds}, \varphi_{qs}, \varphi_{dr}, \varphi_{qr})$ are two-axis components of the spatial vectors of the rotor and stator voltages, currents, and fluxes respectively. The angular synchronous speed of the utility is denoted by ω_s while the electrical angular speed of the rotor is shown by ω_r . Lastly, the number of pole pairs of the machine is defined by P .

2.2. Indirect matrix converter

In this paper, the B2B converter structure is substituted by an IMC to excite the generator. The IMC’s reliability and robustness are more than B2B converter due to removing the bulky electrolytic capacitor. The schematic of an IMC is shown in Figure 2. In the input rectifier section bidirectional switches are used and the fictitious dc-link polarity is denoted by P and N for positive and negative rails. To simplify the calculations, it could be a logical assumption that the switching frequency is very greater than the frequencies of both the output current and input voltage. Based on this assumption, in each cycle of the switching, both the output current and the input voltage could be kept as constants.

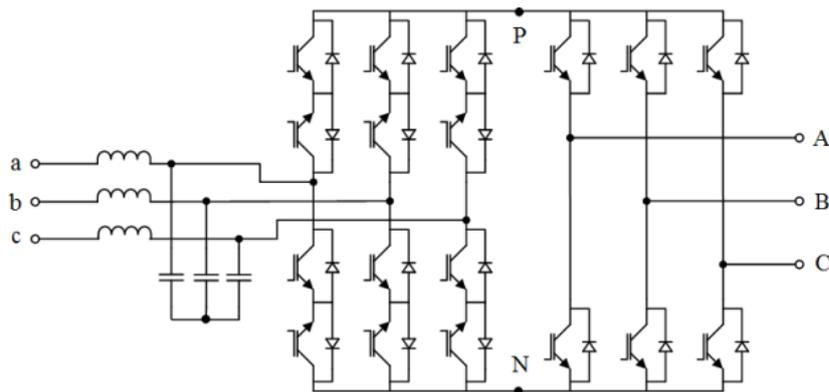


Figure 2. Schematic of a three-phase to three-phase IMC

The rectification part of the IMC must be switched to obtain maximum voltages in the fictitious dc-link and low harmonics of injected currents to the grid. In normal operation mode of the DFIG, the unit power factor is favorable at the rectifier side. To obtain the maximum dc-link voltage at each 60-degree intervals, the input phase voltage with highest absolute value is switched to the positive or negative rails of the fictitious dc-link. Assume the input voltages are as follows:

$$\begin{aligned}
 v_a &= V_p \cos \delta_a = V_p \cos(\omega_i t) \\
 v_b &= V_p \cos \delta_b = V_p \cos(\omega_i t - \frac{2\pi}{3}) \\
 v_c &= V_p \cos \delta_c = V_p \cos(\omega_i t + \frac{2\pi}{3})
 \end{aligned} \tag{2}$$

where V_p is the phase peak of the input voltage. The frequency of the utility is denoted by ω_i . To obtain at the input side unit power factor during normal operation mode, the phase difference between the currents and voltages in the input must be zero. Consequently, the injected currents to the grid are given by:

$$\begin{aligned}
 I_a &= I_p \cos(\omega_i t) \\
 I_b &= I_p \cos(\omega_i t - \frac{2\pi}{3}) \\
 I_c &= I_p \cos(\omega_i t + \frac{2\pi}{3})
 \end{aligned} \tag{3}$$

where I_p is the peak of the currents injected to the grid. It should be noted that on the rectification part of the IMC two switches, one of the top switches and the other from the down switches, are always ON and the four remain switches are OFF.

3. ANFIS-SMC

SMC is a nonlinear variable structure controller. Realization of the control system in this method is simple. In contrast with classical control methods, this scheme does not need the exact mathematical model of the system. The main objective of the wind turbine controller is to decouple controlling of the active and reactive power injecting to the grid. Therefore, the sliding surfaces are considered as (4).

$$\begin{bmatrix} S_d \\ S_q \end{bmatrix} = \begin{bmatrix} I_{drref} - I_{dr} \\ I_{qrref} - I_{qr} \end{bmatrix} \tag{4}$$

Based on the defined surfaces as above, the derivation are as (5).

$$\begin{bmatrix} \dot{S}_d \\ \dot{S}_q \end{bmatrix} = \begin{bmatrix} \dot{I}_{drref} - \dot{I}_{dr} \\ \dot{I}_{qrref} - \dot{I}_{qr} \end{bmatrix} \tag{5}$$

Substituting the derivatives of the rotor active and reactive currents components, we obtain:

$$\begin{bmatrix} \dot{S}_d \\ \dot{S}_q \end{bmatrix} = \begin{bmatrix} I_{drref} - \frac{\alpha}{\sigma L_r} [V_{dr} - R_r \cdot I_{qr} - g \cdot \omega_s \cdot \sigma \cdot L_r \cdot I_{dr} - g \frac{M \cdot V_S}{L_S}] \\ I_{qrref} - \frac{\alpha}{\sigma L_r} [V_{qr} - R_r \cdot I_{dr} + g \cdot \omega_s \cdot \sigma \cdot L_r \cdot I_{qr}] \end{bmatrix} \tag{6}$$

If the K_1 and K_2 functions are defined as follows:

$$\begin{bmatrix} K_1 \\ K_2 \end{bmatrix} = \begin{bmatrix} I_{drref} - \frac{\alpha}{\sigma L_r} [-R_r \cdot I_{qr} - g \cdot \omega_s \cdot \sigma \cdot L_r \cdot I_{dr} - g \frac{M \cdot V_S}{L_S}] \\ I_{qrref} - \frac{\alpha}{\sigma L_r} [-R_r \cdot I_{dr} + g \cdot \omega_s \cdot \sigma \cdot L_r \cdot I_{qr}] \end{bmatrix} \tag{7}$$

then the equation based on these functions and d-q components of the rotor voltages are as (8).

$$\begin{bmatrix} S_d \\ S_q \end{bmatrix} = \begin{bmatrix} \frac{\alpha}{\sigma L_r} V_{dr} + K_1 \\ \frac{\alpha}{\sigma L_r} V_{qr} + K_2 \end{bmatrix} \tag{8}$$

ANFIS-SMC is similar to the classical SMC method that the switching term in controller is substituted with an ANFIS controller shown in Figure 3. Additionally, oscillations of control parameters in ANFIS-SMC controller are much lower than classical SMC. Block diagram of ANFIS controller is presented in Figure 4.

This is the required reactive power of the generator for magnetization of the core. However, the generator is capable of the reactive power injecting to the grid. But, in this study due to operation near the nominal active power the reactive power is set to zero. The currents of the generator are presented in Figure 8. The currents are pure sine waves for the stator and the rotor. The stator currents are with the grid frequency in spite of the rotor currents with the slip frequency.

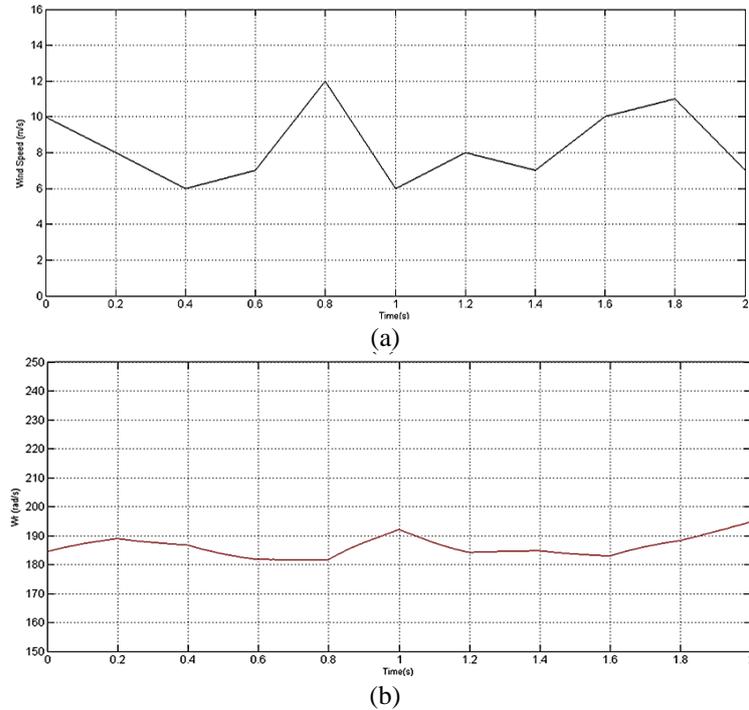


Figure 5. Variable speed operation of the system (a) wind speed profile and (b) rotor speed of the DFIG

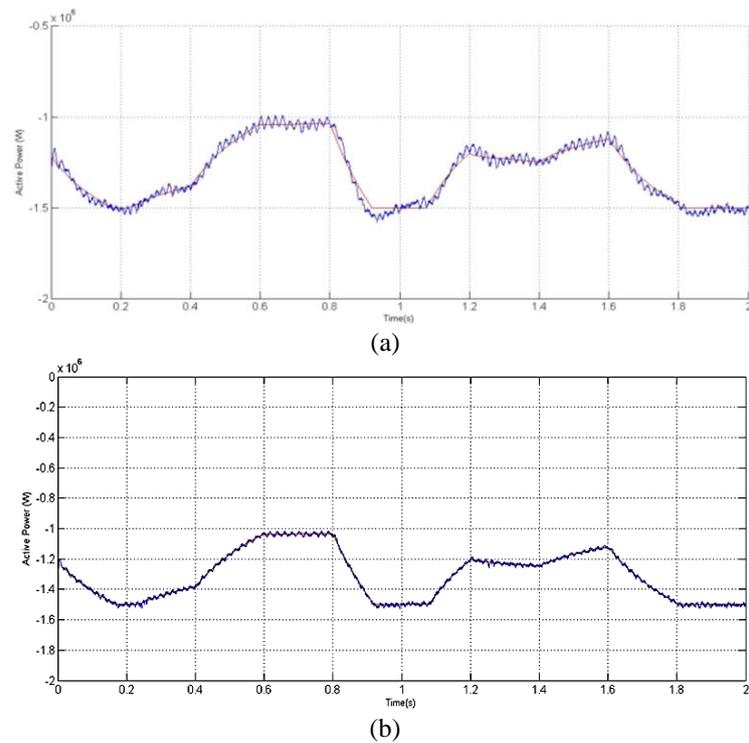


Figure 6. The generator active power injected to the grid: (a) PI controller and (b) ANFIS-SMC controller

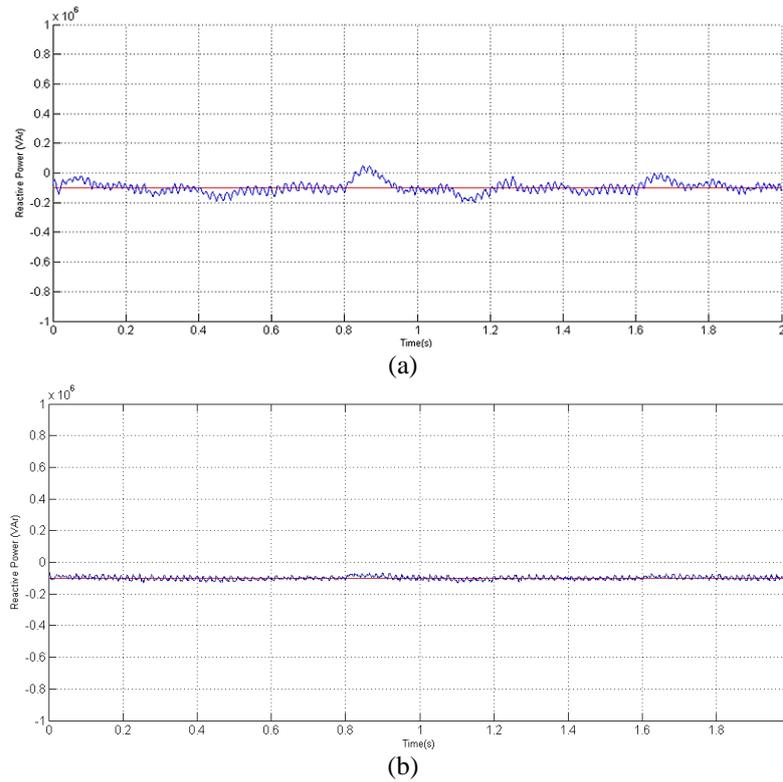


Figure 7. The generator reactive power injected to the grid: (a) PI controller and (b) ANFIS-SMC controller

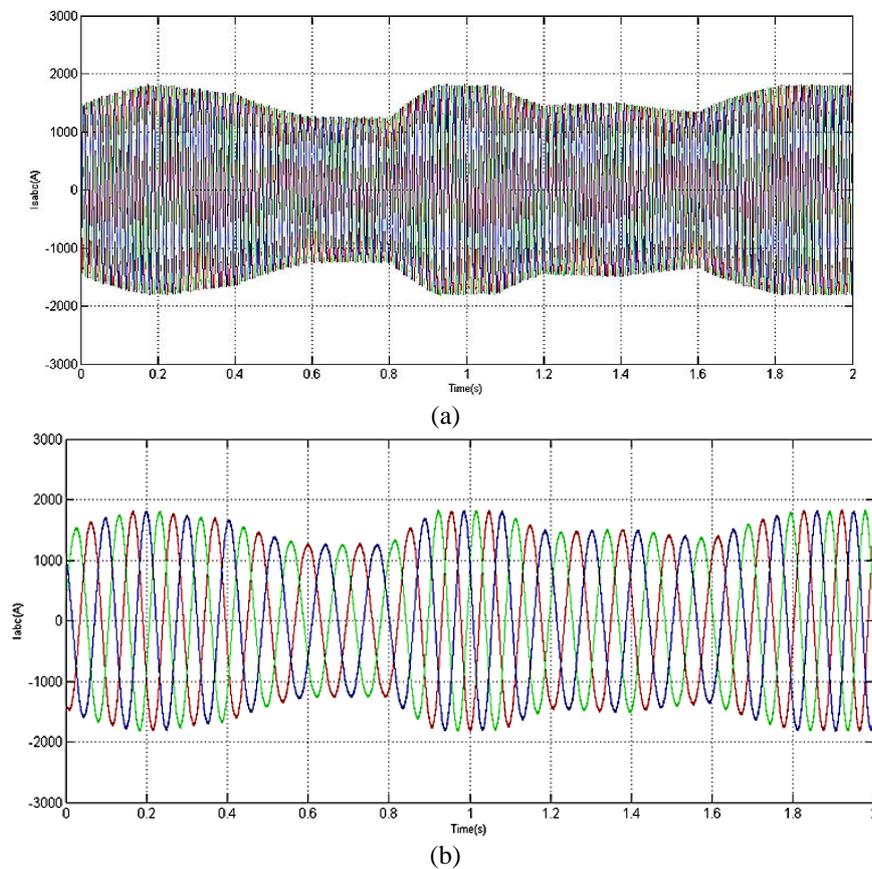


Figure 8. The generator currents: (a) stator currents and (b) rotor currents

5. CONCLUSION

To improve the overall performance and increased reliability of the system, it is proposed in this paper to replace the conventional back-to-back converter with IMC. IMC is a fully silicon-based converter without capacitor that could be used in wind turbines. In this paper a robust ANFIS based sliding mode controller is designed to control the DFIG wind turbine excited by IMC converter. ANFIS is a combination of fuzzy logic control and artificial neural networks. The proposed controller is tested in variable wind speed for two seconds via exact model of switching converter. Simulation results show that in addition to very well reference tracking, the steady state error of the control parameters is reduced about 80 percent. Therefore, the effectiveness of the proposed controller compared to the classical PI controller is verified.

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