Model predictive current control for maximum power point tracking of voltage source inverter based grid connected photovoltaic system

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ABSTRACT

Due to the high demand of grid connected photovoltaic systems, there is a need to track the maximum power point of the PV system. As the output of PV system is dc, there should be a converter, acting as medium between PV system and dc bus capacitor to track maximum power at all the loads. Usually boost converter is acting as medium between PV system and dc link capacitor as the duty cycle of the insulated-gate bipolar transistor in boost converter is in between 0 to 1 for maximum loads during maximum power point tracking (MPPT). To make PV system stable, the balance point is dc bus. If the dc bus voltage is constant, the system will be stable. Then the transfer power will just depend on current. For this purpose, the active current reference signal is to be generated by setting up the reference voltage across dc bus. Here to generate active reference current, PI controller is used and the reference voltage is taken according to the peak voltage of the inverter output voltage. The proposed control strategy was evaluated on a three-phase inverter linked to the grid and supplied by the PV system, which is working under varying irradiation conditions.

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1. INTRODUCTION

The decreasing of conventional power sources and reduction of PV cell cost have increased the using of photovoltaic systems. In photovoltaic (PV) systems maximum power point tracking (MPPT) methods becomes the part of control the power electronic converters [1]. Secondary control algorithms [1]–[5] are used to get better sensitivity and dynamic capability of control algorithms. Usually, renewable energy sources are used in microgrid where the group of loads are interconnected and clearly defined their electrical boundaries and acts as single controllable entity [6], [7] with respect to grid and can be connected in grid mode and islanded mode. for islanded networks there are some sensitivity-based methods to minimize the current during switching action [8], [9]. Usually sliding mode control (SMC) [10] and model predictive control (MPC) [2] algorithm is used to improve dynamic response. Even these techniques demand powerful computing, it has eliminated by technical improvements on micro controllers.

Different algorithms are proposed to track maximum power [11]–[13] from the PV systems at different loads. Adaptive neuro fuzzy inference and P&O MPPT algorithms are combined for better tracking

of the maximum power [14]–[17]. It will give better results than the classical MPPT methods. Mechanical trackers are also used in tracking the maximum power from the PV system [18]–[20], but this is not a practical approach for high power rating PV systems. Advanced control algorithms are used to achieve a better dynamic reaction and sensitive control due to the challenges on the mechanical observer. The main advantage of sliding mode control is its fast dynamic response, robustness and insensitive to system parameters. But this work makes the argument that the model predictive control technique is used to produce the quick dynamic response. Model predictive control method is another method that has similar advantages as SMC. The MPPT algorithm is combined with MPC algorithm [21], [22] to obtain better maximum power tracking performance. Especially this combination gives good performance under variable irradiation conditions. This algorithm manages a Z-source inverter connected to the grid.

In this paper one stage power converter has been used. As two-stage power converters [21], [22] and [23], usage is increasing total cost of the system. But while using single stage power converter [24] series connected PV panel group is needed to get more dc bus voltage. Even there are different solutions to increase dc bus voltage [25]–[28] but one stage power layer with series PV panels is the most preferred model [24]–[29]. In this grid connected single stage power converter, the important parameter is the voltage across dc bus capacitor, which is to be maintained constant irrespective of the load for smooth and proper operation of the Z-source inverter. In this paper boost converter is used as the medium between PV system and dc bus capacitor as the firing angle of the power electronic switch is in between 0 to 1 for maximum loads during MPPT, which is not the case in buck type converter or buck boost converter. In this paper the comparison between buck, boost and buck-boost converter has been given for different load resistance values.

Here boost converter is used as the medium between PV system and dc bus capacitor and to boost up the input voltage of the inverter. To maintain the constant dc bus voltage the active current reference signal is to be generated by setting up the reference voltage across dc bus capacitor and as the reactive current is to be supplied by the grid, the reference reactive current is to be maintained at zero. Here to generate active reference current PI controller is used. And the reference voltage is taken according to the peak voltage of the inverter output voltage. The proposed control strategy was evaluated on a three-phase inverter linked to the grid and supplied by the PV system, which is working under varying irradiation and cloudy conditions.

Here one should think about some operational behavior of PI controller. That is according to irradiation curve, the PV irradiation will be zero for some time and after that it starts increasing. But during the zero-irradiation time, the generated power will be zero, so the converter cannot transfer any power to the dc bus, but the PI controller will work as usual and it will try to make dc bus voltage equal to reference voltage. It cannot be possible because generated power is zero and the PI controller will get saturated, when irradiation increases, the voltage starts increasing, but as already PI controller gets saturated power cannot transfer to the grid through inverter. So, the dc bus voltage reaches to very high value, so there is a need to reset PI controller by designing a reset condition. Here in this work, if the dc bus voltage is greater than 1.05 times that of reference voltage the PI controller will get reset. If there is no saturation point for the direct reference current, the reference current may increase to very higher value which may activate the over current relay and causes to nuisance tripping. So, a saturation point should be provided at the output of the PI controller.

2. SYSTEM MODEL AND OPERATION OF THE PROPOSED SYSTEM

2.1. Operation of the proposed system

Generally, in photo voltaic modules there is one single operating point at any given point in time where maximum power can be drawn, called as maximum power point. In this work MPPT algorithm is used to track the maximum power point of the photo voltaic system by controlling the duty cycle of the boost converter for different load conditions. The generated power from PV system is transferred to dc bus by boost converter and MPPT algorithm. The inverter transfers the power from dc bus to grid. The prediction algorithm is used to generate reference active component of current for inverter. The phase locked loop (PLL) is used to transform the three phase currents into dq0 transformation as the reference current signal is dc. By reducing the control parameter error between the standard and the subsequent sampling interval (Iabc(m+1)), the model predictive control algorithm operates. Model predictive control algorithm can be summarized in two steps as prediction and minimizing. To forecast the control value, discretization techniques are employed. The second stage of the model predictive control algorithm is error minimization. In the cost function and minimization block the PI controller is used to generate switching pulses to inverter through pulse width modulation technique, as shown in the simulation model.

2.2. Modelling of the proposed system

General circuit schema and controlling techniques are shown in Figure 1. PV system is directly connected to the grid through boost converter and 3 phase inverters. The values of filter inductance and capacitance can be calculated by (1) and (2), respectively [30].

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$$L_{filter} = \frac{0.1u_{rms}^2}{2\pi f P_{phase}} \tag{1}$$

$$C_{filter} = \frac{0.05P_{phase}}{2\pi f u_{rms}^2} \tag{2}$$

Where u_{rms} is the line-to-line rms voltage from the grid side and P_{phase} is the single-phase power.

Here PV power is taken as a total system power for theoretical calculations. PV system consist 17 parallel strings and 14 series connected modules per each string and total input power is 50KW. The inductance and capacitance of the boost converter is calculated by (3) and (5), respectively [2].

$$L_{boost} = \frac{(1-D)^2 . D . V_0^2}{2f_{sw} P}$$
(3)

Where D is the duty cycle, is the voltage across dc link capacitor, and f_{sw} is the switching frequency, P is the total power. The duty cycle can be calculated by using the (4).

$$D = 1 - \frac{V_{MPP}}{V_0} \tag{4}$$

Where V_{MPP} is the voltage of the PV system at maximum power point.

$$C_{boost} = \frac{D.P}{V_0^2.\% ripple.f_{sw}}$$
(5)

Here while designing capacitor value, percentage of ripple content to be taken as low as possible, but keep in the mind the cost of the capacitor banks. The maximum power point tracking (MPPT) is used to track maximum energy from PV panels at different irradiance levels. In any MPPT algorithm the duty cycle of the converter is changed to make the load resistance is equal to internal resistance of PV module to transfer maximum power from source to load. The comparison between buck, boost and buck-boost converter is shown in Table 1, in which it was observed that for 90% of the loads the duty cycle value of the boost converter is in between 0 and 1 and the continuous current can be drawn from the PV system due to the input inductance of the boost converter.

The reference current for the inverter is determined using the conventional P&O MPPT algorithm described in [2]. As a result, the model predictive current control approach is the main emphasis of this study, which is utilized to control inverter current (i_{abc}). The general voltage equation for the inverter [31] is given in (6). Here $V_{inverter}$ is the inverter voltage, V_{Lf} is the voltage drop due to filter inductance, V_{Rf} is the voltage drop due to filter resistance and V_{GRID} is the measured grid voltage.

$$V_{inverter} = V_{Lf} + V_{Rf} + V_{GRID} \tag{6}$$

Output current equation can be written by replacing of inductor voltages [31] in (7). Here $i_{(m)}$ is the current sample interval's inverter current, L_f is the filter inductance, $V_{inverter(l)}$ is the inverter voltage vector, $V_{GRID(m)}$ is the current sample interval's measured grid voltage and R_{Lf} is the filter resistance.

$$\frac{di_{(m)}}{dt} = \frac{1}{L_f} \left(V_{inverter(l)} - V_{GRID(m)} - R_{Lf} i_{(m)} \right) \tag{7}$$

Where $V_{GRID(m)}$ is the current sample interval's measured grid voltage. All measured parameters are converted using the Clarke transformation into α - β coordinate. $V_{inverter}$ stands for inverter voltage vector, which is used to make predictions. The voltage space vectors are given in Figure 2. There are two "0" states and six active states in three-phase inverters, as shown in the diagram. It demonstrates that there are eight states in which the inverter can function. An array of phase states represents all voltage states (S_A, S_B, S_C). In (8) – α - β frame [28] can be used to express inverter voltage states. As a result, all of the control parameters in (7) are in the α - β frame.

$$V_{inverter(l)} = \frac{2}{3} \left[S_{A(l)} + (-0.5 + 0.866j) S_{B(l)} + (-0.5 - 0.866j) S_{C(l)} \right]$$
(8)

In the (8), 1 is state number that varies from $\{1, 2, \dots, 8\}$.



Figure 1. General schematic diagram of the system

	Table 1. Du	ty cycle of b	uck, boost and b	uck-boost	converter fo	r different	loads	
$R_0(\Omega)$	δ (boost)	δ (buck)	δ (buck boost)	$R_0(\Omega)$	δ (boost)	δ (buck)	δ (buck boost)	
2	-0.95157	0.51241	0.3388	52	0.61727	2.6128	0.7232	
4	-0.37997	0.72465	0.42017	54	0.62442	2.6626	0.72697	
6	-0.12674	0.88752	0.4702	56	0.63119	2.7114	0.73056	
8	0.024216	1.0248	0.50613	58	0.6376	2.7594	0.734	
10	0.12723	1.1458	0.53397	60	0.64369	2.8066	0.7373	
12	0.20328	1.2551	0.55657	62	0.64949	2.853	0.74046	
14	0.26238	1.3557	0.5755	64	0.65501	2.8986	0.7435	
16	0.31002	1.4493	0.59172	66	0.66028	2.9436	0.74642	
18	0.34948	1.5372	0.60587	68	0.66531	2.9878	0.74924	
20	0.38286	1.6204	0.61838	70	0.67012	3.0314	0.75195	
22	0.41158	1.6995	0.62956	72	0.67474	3.0744	0.75457	
24	0.43663	1.775	0.63964	74	0.67916	3.1169	0.7571	
26	0.45873	1.8475	0.64882	76	0.68341	3.1587	0.75954	
28	0.47842	1.9173	0.65721	78	0.6875	3.2000	0.7619	
30	0.49511	1.9845	0.66494	80	0.69143	3.2408	0.76419	
32	0.51211	2.0496	0.67209	82	0.69522	3.281	0.76641	
34	0.52668	2.1127	0.67874	84	0.69887	3.3208	0.76856	
36	0.54001	2.174	0.68494	86	0.70239	3.3601	0.77065	
38	0.55228	2.2335	0.69074	88	0.70579	3.3989	0.77267	
40	0.56362	2.2916	0.69619	90	0.70908	3.4373	0.77464	
42	0.57413	2.3481	0.70133	92	0.71226	3.4753	0.77655	
44	0.58392	2.4034	0.70618	94	0.71533	3.5129	0.77841	
46	0.59307	2.4574	0.71077	96	0.71832	3.5501	0.78022	
48	0.60164	2.5103	0.71512	100	0.72401	3.6233	0.7837	
50	0 60969	2 562	0 71926					



Figure 2. Space vectors of three-phase two level inverter

3. METHOD FOR MODELING PREDICTIVE CURRENT CONTROL

Between the reference and the next sample period, the MPC algorithm minimizes the control parameter error. MPC consists of two phases. prediction and minimization. Discrete method discretization is employed to anticipate the control parameter. Usually, exact discretization method has high accuracy for low

order and high order systems but it will take a greater number of iterations and more computational time. The forward Euler approximation method also has high accuracy for low order systems with a smaller number of iterations and less computational time. The local truncation error or local discretization error in the Euler method is the error made in approximating the derivative by the difference quotient, whereas the global discretization error at a position is the magnitude of the actual error at the point. so, (9) is a first order differential equation. So, the forward Euler approximation [31] method is preferred.

$$\frac{di}{dt} \approx \frac{i_{(m+1)} - i_{(m)}}{T_s} \tag{9}$$

In (10) can be produced by replacing the derivative in (7) with the Euler technique.

$$\frac{i_{(m+1)}-i_{(m)}}{T_s} = \frac{1}{L_f} \left(V_{inverter} - V_{GRID(m)} - R_{Lf} i_{(m)} \right)$$
(10)

You may get the prediction in (11) by arranging the (10).

$$i_{(m+1)} = \left[\frac{T_s}{L_f} \left(V_{inverter(l)} - V_{GRID(m)} - R_{Lf} i_{(m)} \right) \right] + i_{(m)}$$
(11)

In (11) predicts current values for all possible voltage states (1....n). Error minimization is the second stage of the MPC method, as was already mentioned. Most control methods use the difference between the reference and measured values. Unlike previous approaches, Utilizing the discrepancy between the reference and predicted values, the MPC algorithm. This feature allows the algorithm to create an action for the subsequent phase. As a result, MPC has a greater dynamic range. In MPC algorithms, in order to calculate the error term, cost functions are used. The control algorithm's cost function is represented by (12). Real and imaginary current values are denoted by i_{α} and i_{β} respectively.

$$g = \left| i_{\alpha}^{*} - i_{\alpha(m+1)} \right| + \left| i_{\beta}^{*} - i_{\beta(m+1)} \right|$$
(12)

For each sample interval, the cost function is evaluated along with the prediction equation for all potential voltage vectors. As a result, error values are generated for every possible switching position. The best cost function with the least amount of error is selected during the minimization stage. In order to create switching signals, the optimum cost function is used.

4. SIMULATION RESULTS

Simulation studies are used to verify the system model and proposed control structures in Figure 1. Table 2 lists the simulation parameters. Firstly, the proposed control algorithm's power flow control performance has been tested under various irradiation conditions. The simulation model is shown in Figures 3(a)-(c).

Table 2. System specifications					
Parameter	Value				
<i>u_{rms}</i> (Grid voltage)	380 V				
f (Grid frequency)	50 Hz				
P (Nominal power)	50.72 kW				
L_{Filter} (Filter inductance per phase)	0.0028 H				
<i>L</i> _{boost} (Inductance of boost converter)	1.6236 mH				
C (DC link capacitor)	1000 µF				
PV Panel (17 parallel and 14 series)	50.72 kW _{MPP} ,406 V _{MPP}				
T_s (Sampling time)	10 µs				

Following the dynamic performance analysis, the suggested control algorithm's power flow control performance was tested under various irradiation situations. Figure 4 shows the irradiation curve during the course of 30 seconds [11]. This paper presents a method that overcomes the problem of the confusion during fast irradiation change in the classical MPPT as well as in model predictive control (MPC) based MPPTs available in the literature.



Figure 3. Simulation models for grid connected PV system: (a) DC side of the system, (b) AC side of the system, and (c) reference active current generation and control unit



Figure 4. Irradiation curve

Figure 5 shows the PV power results. The MPPT algorithm tracks the MPP with 98 percent efficiency, as shown in Figures 5(a) and 5(b). Even though efficiency decreased in cloudy conditions, after these times, efficiency increased to 98 percent. Figures 5(c) and 5(d) show the voltage and current results for the PV panel group, respectively. While the current rises to 125 A, The PV voltage is fairly stable. It results from the nature of MPPT control in PVs and offers information on control effectiveness. The reference voltage for the PI controller to generate direct component of reference current, should be selected based on the voltage across dc bus capacitor, when it is disconnected from the PV system and charged by the inverter. As the inverter is mostly buck type inverter, The dc bus voltage needs to be higher than the inverter output voltage's maximum value.

Here one should concentrate on some operational behavior of PI controller. That is according to irradiation curve, the PV irradiation will be zero for some time and after that it starts increasing. But during the zero-irradiation time, the generated power will be zero, so the converter cannot transfer any power to the dc bus, but the PI controller will work as usual and it will try to make dc bus voltage equal to reference voltage. It cannot be possible because generated power is zero and the PI controller will get saturated, when irradiation increases, the voltage starts increasing, but as already PI controller gets saturated power cannot transfer to the grid through inverter. So, the dc bus voltage reaches to very high value, so there is a need to reset PI controller by designing a reset condition.



Figure 5. PV results: (a) power, (b) MPPT efficiency, (c) PV voltage, and (d) PV current

Here in this work, if the dc bus voltage is greater than 1.05 times that of reference voltage the PI controller will get reset. If there is no saturation point for the direct reference current, the reference current may increase to very higher value which may activate the over current relay and causes to nuisance tripping. So, a saturation point should be provided at the output of the PI controller to limit the current with in specified range. Figure 6 shows the reference voltage that must be selected for proper operation of PI controller to generate active component of reference current. The active reference current will be generated based on the difference between dc bus voltage and reference voltage. The active component of the reference current is shown in Figure 7. And the voltage across the capacitor, which is also called as dc bus voltage is maintained constant, irrespective of the load and for the given system it is maintained at 800 V as shown in Figure 8.





Figure 6. Reference voltage for dc bus voltage

Figure 7. Reference current for the PI controller

Aside from the MPPT algorithm's performance, one must examine the power, effectiveness, and harmonic distortions of the inverter side. The MPC method correctly fixes the inverter currents to the

reference value (Iref*) as illustrated in Figure 9(a). The inverter's output currents are also altered in accordance with the irradiation curve in Figure 4. Because The reference current value is adjusted by the MPPT algorithm and MPC. Harmonics are less than 5% for active power values greater than 20 kW, as shown in Figure 9(b). The THD value is less than the limit set forth in the standards [32].

Here the PI controller will trigger the inverter to make the peak current of the inverter output equal to the active component of the reference current as shown in Figure 10. Detailed results of inverter currents shown in Figure 11(a). The MPC algorithm successfully regulates the inverter currents to the reference. According to recent tracking data, the combined control algorithms reliably follow the MPP under hazy and variable irradiance conditions. Detailed results of the inverter voltages are shown in Figure 11(b). The inverter voltages are successfully synchronized with the grid voltage of line-to-line RMS value of 380 V. Irrespective of the load the MPC algorithm always keep the inverter voltage equals to the grid voltage.



Figure 9. Inverter results: (a) currents and (b) THD



Figure 10. Line currents of the inverter and reference active current



Figure 11. Detailed inverter results: (a) line currents and (b) line voltages

To evaluate the total efficiency of the proposed system, the transferred power to grid at the output of the inverter was compared to ideal MPP curves. Figure 12 depicts the power losses between the inverter output power and the desired maximum power curve. Around the nominal power, power losses are quite high. Filter inductance

is to blame. The losses in a practical inductor come mainly from the parasitic equivalent resistance (which is practically the ohmic resistance of the windings which depends on the geometry of the wire used in winding the inductor as well as the frequency) and the core losses. To reduce the major losses in the filter inductor the toroidal shape inductor is used. Due to its symmetry, the toroidal shape has the benefit of having a low leakage flux, or the quantity of magnetic flux that escapes the core is low, therefore the core losses will be less and by keeping the ohmic resistance of the inductor low the copper losses can be reduced. Even if losses increased, figure demonstrates that for power values more than 5 kW, the minimum system efficiency is around 90%.



Figure 12. Inverter output power and ideal MPP curve

5. CONCLUSION

A combined control algorithm for PV systems is suggested in the paper. The MPPT and MPC algorithms are used in the suggested control strategy. The integrated algorithm provided efficient tracking under variable irradiance and foggy situations because of the quick and sensitive control capabilities of MPC. In addition to the combined algorithm performance testing, the step-change tests demonstrate that the MPC algorithm has a superior capability for current control. It has been done to analyze the efficiency of both control and power structures. The MPPT method's control efficiency is 98 percent, while the inverter efficiency is 92 percent. Both the inverter and the control algorithms overall system efficiency are around 90%.

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